## TOSHIBA CMOS Digital Integrated Circuit Silicon Monolithic

## T6L79

## Source Driver for TFT LCD Panels

The T6L79 is a 64 -grayscale-level and 384 -channel output source driver for TFT LCD panels. The device accepts 8 -bit $\times$ 6 -dot digital CMOS-level inputs, for which the direction of data transfer can be selected by the U/D pin. The $10(5 \times 2)$ external power supply and the internal DA converter realize display of 260,000 colors, on which reference analog voltage inputs is made.

The DA converter supports one side dot line inversion, achieving high picture quality. The output dynamic range is a generous 7.3 to $9.8 \mathrm{Vp}-\mathrm{p}$.

Base on high-speed CMOS, the T6L79 offers both low power consumption and high-speed operation. To configure an SXGA or XGA-compatible TFT-LCD module, it allows a maximum operating frequency 45 MHz .


## Features

- Grayscale data : Digital CMOS-level 36-bit (6-bit $\times 6$-input) parallel transfer method, selectable transfer direction
- Panel drive outputs
: 384 outputs, 64 grayscale levels, R -DAC system, reference analog voltage inputs, 10 external power supplies ( $5 \times 2$ ), one side dot line inversion drive
- High-speed operation : 45 MHz max
- Power supply voltage : Digital power supply voltage . . . 3.0 to 3.6 V Analog power supply voltage . . . 7.5 to 10.0 V
- Operating temperature: -20 to $75^{\circ} \mathrm{C}$
- Package: Tape carrier package (TCP)
- Cascading of multiple devices


## Block Diagram



## Pin Assignment



The figure above shows an example of the pin assignment in the TCP. It does not specify the pad layout on the chip. For the latest TCP specifications, contact Toshiba or your local sales office.

## Pin Functions

| Pin Name | I/O | Function |
| :---: | :---: | :---: |
| $\begin{aligned} & \mathrm{DI} / \mathrm{O} \\ & \mathrm{DO} / \mathrm{I} \end{aligned}$ | I/O | Data transfer enable pins <br> These pins are used to input/output grayscale data. <br> Input and output are switched as shown below according to the setting of the U/D pin. <br> When DI/O or DO/I is set to input <br> High level input to the pin is latched into the internal logic in sync with the rising edge of CPH. If the internal logic is in standby state, the device is ready for data transfer. Grayscale data are sequentially latched starting from the next rising edge of CPH. <br> When $\mathrm{DI} / \mathrm{O}$ or DO/I is set to output <br> Sends enable signal to the T6L79 at the next stage of the LCD driver. After outputting High level, the pin enters standby state. |
| U/D | I | Transfer direction select pin <br> Controls the transfer direction of grayscale data. The data are transferred in the following order in sync with the rising edge of CPH: <br> When U/D = High: OUT1 to OUT6, OUT7 to OUT12 . . . OUT379 to OUT384 <br> When U/D = Low: OUT379 to OUT384, OUT373 to OUT378 . . . OUT1 to OUT6 <br> The voltage applied to this pin must be DC level, High or Low. |
| CPH | I | Data transfer clock pin <br> Transfers grayscale data. Sequentially latches grayscale data into REG1 in sync with the rising edge of CPH. Input clocks at least three cycles after High level of the LOAD pin. |
| D00 to D05 <br> D10 to D15 <br> D20 to D25 <br> D30 to D35 <br> D40 to D45 <br> D50 to D55 | I | Grayscale data bus pins <br> Inputs output data consisting of 6 bits for six channels in one transfer. The relationship between grayscale data and output pins is as follows: <br> Grayscale data $=32 \times \mathrm{Dn} 5+16 \times \mathrm{Dn} 4+8 \times \mathrm{Dn} 3+4 \times \mathrm{Dn} 2+2 \times \mathrm{Dn} 1+\mathrm{Dn} 0$ <br> (*) $\mathrm{n}=0,1, \ldots, 5$ |
| LOAD | I | Data load input pin <br> High level input to the pin is latched into the internal logic in sync with the rising edge of CPH. Then, the REG1 data are transferred to REG2. The voltage changes to that corresponding to grayscale data in sync with the falling edge of LOAD. |
| DINV | I | Data polarity inverting pin <br> Selects inversion or non-inversion of grayscale data bus at the rising edge of CPH. <br> DINV = Low: Data bus not inverted <br> DINV = High: Data bus inverted <br> " H " or "L" level is discriminated at the rising edge of CPH same as gray data bus. |
| POL | I | Polarity inverting pin <br> Latches the input signal into the internal logic in sync with the rising edge of LOAD. $\begin{aligned} & \text { POL = Low: The reference voltage for odd-numbered outputs is input from V0 to V4; } \\ & \text { for even-numbered outputs, V5 to V9. } \\ & \text { POL = High: } \begin{array}{l} \text { The reference voltage for odd-numbered outputs is input from V5 to V9; } \\ \text { for even-numbered outputs, V0 to V4. } \end{array} \end{aligned}$ |
| V0 to V9 | I | Reference analog voltage input pins <br> Externally input reference analog voltage inputs voltage. <br> Hold the input voltage during output of the voltage corresponding to grayscale data. $\begin{aligned} & \mathrm{AV} \text { DD }-0.1>\mathrm{V} 0>\mathrm{V} 1>\mathrm{V} 2>\mathrm{V} 3>\mathrm{V} 4>0.5 \times \mathrm{AV} \text { DD }>\mathrm{V} 5>\mathrm{V} 6>\mathrm{V} 7>\mathrm{V} 8> \\ & \mathrm{V} 9>\mathrm{AV}_{\mathrm{SS}}+0.1 \end{aligned}$ |
| TEST | I | Test pin <br> Leave the pin open or set to $D V_{\text {SS }}$ level. |
| $\overline{\text { TESTP }}$ | I | Test pin <br> Controlling the output amp constant current supply reduces current dissipation. <br> In Low Power Mode ( TESTP $=$ Low), current dissipation is reduced to $2 / 3$ of the value in Normal Current Dissipation Mode. The pin is internally pulled up to DVDD. $\begin{aligned} & \overline{\text { TESTP }}=\text { High or the pin left open: Normal Current Dissipation Mode } \\ & \overline{\text { TESTP }}=\text { Low: Low-Current-Dissipation Mode } \end{aligned}$ |
| OUT1 to OUT384 | 0 | LCD panel drive pins |
| $A V_{D D}$ |  | Analog power supply pin |
| $\mathrm{AV}_{\text {SS }}$ |  | Analog GND pin Apply the same voltage as that of digital GND pin. |
| DV ${ }_{\text {DD }}$ |  | Digital power supply pin |
| DV ${ }_{\text {SS }}$ |  | Digital GND pin <br> Apply the same voltage as that of analog GND pin. |

## Device Operation

## (1) Starting data transfer

High level input to the data transfer enable pin (DI/O or DO/I) is latched into the internal logic in sync with the rising edge of CPH , setting the device to ready for transfer.

Set the period for High level input to the data transfer enable pin to one clock.

## (2) Data transfer method

Grayscale data are transferred from the data bus to the sampling register (REG1) in sync with the rising edge of CPH. Grayscale data for six channels are simultaneously written in one transfer. Transfer completes after 64 times and the device enters standby state. Data to be written to REG1 are the result of operations between the grayscale data bus and DINV.

Note 1: Do not input High level to LOAD during data transfer.

## (3) Ending data transfer

High level is output from the data transfer enable pin (DO/I or DI/O) from the rising edge of CPH , one cycle before the last data are latched, to the next rising edge of CPH.

## (4) LCD panel drive output

Inputting High level to LOAD in sync with the rising edge of CPH following High level output from DO/I (DI/O) transfers data from the sampling register (REG1) to the load register (REG2) and outputs a voltage corresponding to the grayscale data in sync with the falling edge of LOAD.

Note 2: Input High level to LOAD at least three clock cycles.

The T6L79 has a capacity for polarity inverting corresponding to dot-inverting operation. By POL signal to the pin which is latched into the internal logic in sync with the rising edge of LOAD, the polarity of odd-numbered outputs and even-numbered outputs are inverted.

By changing the cycle of POL signal, $n$-line inverting operation is available to correspond.

| POL = "L" | The reference voltage for odd-numbered outputs is input <br> from V0 to V4, for even-numbered outputs, V5 to V9 |
| :---: | :--- |
| POL = "H" | The reference voltage for odd-numbered outputs is input <br> from V5 to V9, for even-numbered outputs, V0 to V4 |

## (5) TESTP (Low power control function)

Output amp bias current can be switched between two values using the $\overline{\text { TESTP }}$ pin. The current dissipation in Low Current Dissipation Mode can be reduced to $2 / 3$ of that in Normal Current Dissipation mode.

Note 3: Input the stabilized DC voltage (DVDD, $\mathrm{DV}_{\mathrm{SS}}$ ) to the $\overline{\mathrm{TESTP}}$ pin.

| $\overline{\text { TESTP }}=$ "H" | $\overline{\text { TESTP }}=" \mathrm{~L} "$ |
| :---: | :---: |
| Normal current dissipation mode | Low current dissipation mode |

## (6) Reference analog voltage inputs power supply circuit

The DA converter consists of ladder resistors and switches. Resistors are serially connected between the supply voltage input pins for reference analog voltage inputs.


- Resistors between power supply pins for reference analog voltage inputs (typical)

| Resistor <br> Name | Resistance <br> Values | Resistor <br> Name | Resistance <br> Values | Resistor <br> Name | Resistance <br> Values | Resistor <br> Name | Resistance <br> Values |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $R_{0}$ | 648.0 | $R_{16}$ | 243.0 | $R_{32}$ | 81.0 | $R_{48}$ | 81.0 |
| $R_{1}$ | 607.5 | $R_{17}$ | 202.5 | $R_{33}$ | 81.0 | $R_{49}$ | 81.0 |
| $R_{2}$ | 567.0 | $R_{18}$ | 202.5 | $R_{34}$ | 81.0 | $R_{50}$ | 81.0 |
| $R_{3}$ | 526.5 | $R_{19}$ | 202.5 | $R_{35}$ | 81.0 | $R_{51}$ | 81.0 |
| $R_{4}$ | 486.0 | $R_{20}$ | 162.0 | $R_{36}$ | 81.0 | $R_{52}$ | 81.0 |
| $R_{5}$ | 445.5 | $R_{21}$ | 162.0 | $R_{37}$ | 81.0 | $R_{53}$ | 121.5 |
| $R_{6}$ | 445.5 | $R_{22}$ | 162.0 | $R_{38}$ | 81.0 | $R_{54}$ | 121.5 |
| $R_{7}$ | 405.0 | $R_{23}$ | 121.5 | $R_{39}$ | 81.0 | $R_{55}$ | 121.5 |
| $R_{8}$ | 405.0 | $R_{24}$ | 121.5 | $R_{40}$ | 81.0 | $R_{56}$ | 162.0 |
| $R_{9}$ | 324.0 | $R_{25}$ | 121.5 | $R_{41}$ | 81.0 | $R_{57}$ | 162.0 |
| $R_{10}$ | 324.0 | $R_{26}$ | 121.5 | $R_{42}$ | 81.0 | $R_{58}$ | 202.5 |
| $R_{11}$ | 283.5 | $R_{27}$ | 81.0 | $R_{43}$ | 81.0 | $R_{59}$ | 202.5 |
| $R_{12}$ | 283.5 | $R_{28}$ | 81.0 | $R_{44}$ | 81.0 | $R_{60}$ | 243.0 |
| $R_{13}$ | 283.5 | $R_{29}$ | 81.0 | $R_{45}$ | 81.0 | $R_{61}$ | 405.0 |
| $R_{14}$ | 243.0 | $R_{30}$ | 81.0 | $R_{46}$ | 81.0 | $R_{62}$ | 648.0 |
| $R_{15}$ | 243.0 | $R_{31}$ | 81.0 | $R_{47}$ | 81.0 |  |  |

## (7) Relationship between grayscale data and output voltage

The output voltage is determined according to the grayscale data values and the ten power supply voltages for reference analog voltage inputs. When U/D is High or Low, the grayscale data and the output pins correspond as shown in the table below.

| Grayscale Data | Output |  |
| :---: | :---: | :---: |
|  | $\mathrm{U} / \mathrm{D}=$ "H" | $\mathrm{U} / \mathrm{D}=$ "L" |
| D00 to D05 | OUT $(6 m-5)$ | OUT $(6 p-5)$ |
| D10 to D15 | OUT $(6 m-4)$ | OUT $(6 p-4)$ |
| D20 to D25 | OUT $(6 m-3)$ | OUT $(6 p-3)$ |
| D30 to D35 | OUT $(6 m-2)$ | OUT $(6 p-2)$ |
| D40 to D45 | OUT $(6 m-1)$ | OUT $(6 p-1)$ |
| D50 to D55 | OUT $(6 m)$ | OUT $(6 p)$ |

(*) $m=1,2,3, \cdots, 62,63,64$
$p=64,63,62, \cdots, 3,2,1$

- Schematic of reference analog voltage inputs

- Relationship 1 between grayscale data and output voltage (grayscale data: 00 H to 1 FH )
(*) $n=0,1, \cdots, 5$

| Grayscale Data | Dn5 Dn4 Dn3 Dn2 Dn1 Dn0 |  |  |  |  |  | Output Voltage (reference value) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  | $\mathrm{V} 5<\mathrm{V}$ | $\begin{gathered} \overline{\mathrm{V} 4}<\mathrm{V} 3<\mathrm{V} 2<\mathrm{V} 1<\mathrm{V} 0<\mathrm{AV}_{\mathrm{DD}}, \\ \text { DINV }=\mathrm{L} \end{gathered}$ | $\mathrm{AV}_{\mathrm{SS}}$ | $\begin{gathered} <\mathrm{V} 9<\mathrm{V} 8<\mathrm{V} 7<\mathrm{V} 6<\mathrm{V} 5<\mathrm{V} 4, \\ \text { DINV }=\mathrm{L} \end{gathered}$ |
| 00H | 0 | 0 | 0 | 0 | 0 | 0 | VOOH | Vo | V00H' | V9 |
| 01H | 0 | 0 | 0 | 0 | 0 | 1 | V01H | V1 + (V0 - V1) $\times 5872.5 / 6520.5$ | V01H' | V9 + (V8-V9) × 648.0/6520.5 |
| 02H | 0 | 0 | 0 | 0 | 1 | 0 | V02H | V1 + (V0 - V1) $\times 5265.0 / 6520.5$ | V02H' | V9 + (V8-V9) $\times 1255.5 / 6520.5$ |
| 03H | 0 | 0 | 0 | 0 | 1 | 1 | V03H | V1 + (V0 - V1) $\times 4698.0 / 6520.5$ | V03H' | V9 + (V8-V9) $\times 1822.5 / 6520.5$ |
| 04H | 0 | 0 | 0 | 1 | 0 | 0 | V04H | V1 + (V0 - V1) $\times 4171.5 / 6520.5$ | V04H' | V9 + (V8-V9) $\times 2349.0 / 6520.5$ |
| 05H | 0 | 0 | 0 | 1 | 0 | 1 | V05H | V1 + (V0 - V1) $\times 3685.5 / 6520.5$ | V05H' | V9 + (V8-V9) $\times 2835.0 / 6520.5$ |
| 06H | 0 | 0 | 0 | 1 | 1 | 0 | V06H | V1 + (V0 - V1) $\times 3240.0 / 6520.5$ | V06H' | V9 + (V8-V9) $\times 3280.5 / 6520.5$ |
| 07H | 0 | 0 | 0 | 1 | 1 | 1 | V07H | V1 + (V0 - V1) $\times 2794.5 / 6520.5$ | V07H' | V9 + (V8-V9) $\times 3726.0 / 6520.5$ |
| 08H | 0 | 0 | 1 | 0 | 0 | 0 | V08H | V1 + (V0 - V1) $\times 2389.5 / 6520.5$ | V08H' | V9 + (V8-V9) $\times 4131.0 / 6520.5$ |
| 09H | 0 | 0 | 1 | 0 | 0 | 1 | V09H | V1 + (V0 - V1) $\times 1984.5 / 6520.5$ | V09H' | V9 + (V8-V9) $\times 4536.0 / 6520.5$ |
| 0AH | 0 | 0 | 1 | 0 | 1 | 0 | VOAH | V1 + (V0 - V1) $\times 1660.5 / 6520.5$ | VOAH' | V9 + (V8 - V9) × 4860.0/6520.5 |
| OBH | 0 | 0 | 1 | 0 | 1 | 1 | VOBH | V1 + (V0 - V1) $\times 1336.5 / 6520.5$ | VOBH' | V9 + (V8 - V9) × 5184.0/6520.5 |
| OCH | 0 | 0 | 1 | 1 | 0 | 0 | VOCH | V1 + (V0 - V1) $\times 1053.0 / 6520.5$ | VOCH' | V9 + (V8 - V9) × 5467.5/6520.5 |
| ODH | 0 | 0 | 1 | 1 | 0 | 1 | VODH | V1 + (V0 - V1) $\times 769.5 / 6520.5$ | VODH' | V9 + (V8 - V9) × 5751.0/6520.5 |
| OEH | 0 | 0 | 1 | 1 | 1 | 0 | VOEH | $\mathrm{V} 1+(\mathrm{V} 0-\mathrm{V} 1) \times 486.0 / 6520.5$ | VOEH' | V9 + (V8 - V9) × 6034.5/6520.5 |
| OFH | 0 | 0 | 1 | 1 | 1 | 1 | VOFH | V1 + (V0 - V1) $\times 243.0 / 6520.5$ | VOFH' | V9 + (V8 - V9) × 6277.5/6520.5 |
| 10H | 0 | 1 | 0 | 0 | 0 | 0 | V10H | V1 | V10H' | V8 |
| 11H | 0 | 1 | 0 | 0 | 0 | 1 | V11H | V2 + (V1 - V2) × 1984.5/2227.5 | V11H' | V8 + (V7 - V8) $\times 243.0 / 2227.5$ |
| 12H | 0 | 1 | 0 | 0 | 1 | 0 | V12H | V2 + (V1 - V2) $\times 1782.0 / 2227.5$ | V12H' | V8 + (V7 - V8) $\times 445.5 / 2227.5$ |
| 13H | 0 | 1 | 0 | 0 | 1 | 1 | V13H | V2 + (V1 - V2) $\times 1579.5 / 2227.5$ | V13H' | $\mathrm{V} 8+(\mathrm{V} 7-\mathrm{V} 8) \times 648.0 / 2227.5$ |
| 14H | 0 | 1 | 0 | 1 | 0 | 0 | V14H | V2 + (V1 - V2) $\times 1377.0 / 2227.5$ | V14H' | V8 + (V7 - V8) $\times 850.5 / 2227.5$ |
| 15H | 0 | 1 | 0 | 1 | 0 | 1 | V15H | V2 + (V1 - V2) $\times 1215.0 / 2227.5$ | V15H' | V8 + (V7 - V8) $\times 1012.5 / 2227.5$ |
| 16H | 0 | 1 | 0 | 1 | 1 | 0 | V16H | V2 + (V1 - V2) $\times 1053.0 / 2227.5$ | V16H' | V8 + (V7 - V8) $\times 1174.5 / 2227.5$ |
| 17H | 0 | 1 | 0 | 1 | 1 | 1 | V17H | V2 + (V1 - V2) × 891.0/2227.5 | V17H' | V8 + (V7 - V8) $\times 1336.5 / 2227.5$ |
| 18H | 0 | 1 | 1 | 0 | 0 | 0 | V18H | V2 + (V1 - V2) × 769.5/2227.5 | V18H' | V8 + (V7 - V8) $\times 1458.0 / 2227.5$ |
| 19H | 0 | 1 | 1 | 0 | 0 | 1 | V19H | V2 + (V1 - V2) × 648.0/2227.5 | V19H' | V8 + (V7 - V8) $\times 1579.5 / 2227.5$ |
| 1 AH | 0 | 1 | 1 | 0 | 1 | 0 | V1AH | $\mathrm{V} 2+(\mathrm{V} 1-\mathrm{V} 2) \times 526.5 / 2227.5$ | V1AH' | V8 + (V7 - V8) $\times 1701.0 / 2227.5$ |
| 1BH | 0 | 1 | 1 | 0 | 1 | 1 | V1BH | $\mathrm{V} 2+(\mathrm{V} 1-\mathrm{V} 2) \times 405.0 / 2227.5$ | V1BH' | V8 + (V7 - V8) $\times 1822.5 / 2227.5$ |
| 1 CH | 0 | 1 | 1 | 1 | 0 | 0 | V1CH | $\mathrm{V} 2+(\mathrm{V} 1-\mathrm{V} 2) \times 324.0 / 2227.5$ | V1CH' | V8 + (V7 - V8) $\times 1903.5 / 2227.5$ |
| 1DH | 0 | 1 | 1 | 1 | 0 | 1 | V1DH | $\mathrm{V} 2+(\mathrm{V} 1-\mathrm{V} 2) \times 243.0 / 2227.5$ | V1DH' | V8 + (V7 - V8) $\times 1984.5 / 2227.5$ |
| 1EH | 0 | 1 | 1 | 1 | 1 | 0 | V1EH | $\mathrm{V} 2+(\mathrm{V} 1-\mathrm{V} 2) \times 162.0 / 2227.5$ | V1EH' | V8 + (V7 - V8) $\times 2065.5 / 2227.5$ |
| 1FH | 0 | 1 | 1 | 1 | 1 | 1 | V1FH | V2 + (V1 - V2) $\times 81.0 / 2227.5$ | V1FH' | V8 + (V7 - V8) $\times 2146.5 / 2227.5$ |

## - Relationship 2 between grayscale data and output voltage

 (grayscale data: 20H to 3FH)(*) $n=0,1, \cdots, 5$

| Grayscale Data |  |  | Dn3 | Dn2 | Dn1 | Dn0 | Output Voltage (reference value) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  | $\begin{gathered} \mathrm{V} 4<\mathrm{V} 3<\mathrm{V} 2<\mathrm{V} 1<\mathrm{V} 0<\mathrm{AV}_{\mathrm{DD}}, \\ \text { DINV }=\mathrm{L} \end{gathered}$ | $\mathrm{AV}_{\mathrm{SS}}$ | $\begin{gathered} <\mathrm{V} 9<\mathrm{V} 8<\mathrm{V} 7<\mathrm{V} 6<\mathrm{V} 5<\mathrm{V} 4, \\ \text { DINV }=\mathrm{L} \end{gathered}$ |
| 20H | 1 | 0 | 0 | 0 | 0 | 0 | 20H | V2 | 20H' | V7 |
| 21H | 1 | 0 | 0 | 0 | 0 | 1 | 21H | V3 + (V2 - V3) × 1215.0/1296.0 | 21H' | V7 + (V6-V7) $\times$ 81.0/1296.0 |
| 22 H | 1 | 0 | 0 | 0 | 1 | 0 | 22H | V3 + (V2 - V3) × 1134.0/1296.0 | 22H' | V7 + (V6 - V7) × 162.0/1296.0 |
| 23H | 1 | 0 | 0 | 0 | 1 | 1 | 23H | V3 + (V2-V3) $\times 1053.0 / 1296.0$ | 23H' | V7 + (V6-V7) × 243.0/1296.0 |
| 24H | 1 | 0 | 0 | 1 | 0 | 0 | 24H | V3 + (V2 - V3) $\times 972.0 / 1296.0$ | 24H' | V7 + (V6-V7) × 324.0/1296.0 |
| 25H | 1 | 0 | 0 | 1 | 0 | 1 | 25H | V3 + (V2 - V3) × 891.0/1296.0 | 25H' | V7 + (V6 - V7) × 405.0/1296.0 |
| 26H | 1 | 0 | 0 | 1 | 1 | 0 | 26H | V3 + (V2-V3) $\times 810.0 / 1296.0$ | 26H' | V7 + (V6 - V7) $\times 486.0 / 1296.0$ |
| 27H | 1 | 0 | 0 | 1 | 1 | 1 | 27H | V3 + (V2-V3) × 729.0/1296.0 | 27H' | V7 + (V6-V7) $\times 567.0 / 1296.0$ |
| 28H | 1 | 0 | 1 | 0 | 0 | 0 | 28H | V3 + (V2-V3) × 648.0/1296.0 | 28H' | V7 + (V6 - V7) × 648.0/1296.0 |
| 29H | 1 | 0 | 1 | 0 | 0 | 1 | 29H | V3 + (V2-V3) $\times 567.0 / 1296.0$ | 29H' | V7 + (V6-V7) $\times 729.0 / 1296.0$ |
| 2AH | 1 | 0 | 1 | 0 | 1 | 0 | 2AH | V3 + (V2-V3) $\times 486.0 / 1296.0$ | 2AH' | V7 + (V6-V7) $\times$ 810.0/1296.0 |
| 2BH | 1 | 0 | 1 | 0 | 1 | 1 | 2BH | V3 + (V2 - V3) $\times 405.0 / 1296.0$ | 2BH' | V7 + (V6-V7) × 891.0/1296.0 |
| 2CH | 1 | 0 | 1 | 1 | 0 | 0 | 2CH | V3 + (V2-V3) $\times$ 324.0/1296.0 | 2CH' | V7 + (V6 - V7) $\times$ 972.0/1296.0 |
| 2DH | 1 | 0 | 1 | 1 | 0 | 1 | 2DH | V3 + (V2 - V3) $\times 243.0 / 1296.0$ | 2DH' | V7 + (V6 - V7) $\times 1053.0 / 1296.0$ |
| 2EH | 1 | 0 | 1 | 1 | 1 | 0 | 2EH | V3 + (V2 - V3) × 162.0/1296.0 | 2EH' | V7 + (V6 - V7) $\times 1134.0 / 1296.0$ |
| 2 FH | 1 | 0 | 1 | 1 | 1 | 1 | 2FH | V3 + (V2 - V3) × 81.0/1296.0 | 2FH' | V7 + (V6 - V7) $\times$ 1215.0/1296.0 |
| 30H | 1 | 1 | 0 | 0 | 0 | 0 | 30H | V3 | 30H' | V6 |
| 31 H | 1 | 1 | 0 | 0 | 0 | 1 | 31H | V4 + (V3 - V4) $\times 2713.5 / 2794.5$ | 31H' | V6 + (V5-V6) × 81.0/2794.5 |
| 32 H | 1 | 1 | 0 | 0 | 1 | 0 | 32H | V4 + (V3 - V4) $\times 2632.5 / 2794.5$ | 32H' | V6 + (V5-V6) × 162.0/2794.5 |
| 33H | 1 | 1 | 0 | 0 | 1 | 1 | 33H | V4 + (V3 - V4) $\times 2551.5 / 2794.5$ | 33H' | V6 + (V5-V6) × 243.0/2794.5 |
| 34H | 1 | 1 | 0 | 1 | 0 | 0 | 34H | V4 + (V3 - V4) $\times 2470.5 / 2794.5$ | 34H' | V6 + (V5-V6) $\times$ 324.0/2794.5 |
| 35H | 1 | 1 | 0 | 1 | 0 | 1 | 35H | V4 + (V3 - V4) $\times 2389.5 / 2794.5$ | 35H' | V6 + (V5-V6) $\times 405.0 / 2794.5$ |
| 36H | 1 | 1 | 0 | 1 | 1 | 0 | 36H | V4 + (V3 - V4) $\times 2268.0 / 2794.5$ | 36H' | V6 + (V5-V6) × 526.5/2794.5 |
| 37H | 1 | 1 | 0 | 1 | 1 | 1 | 37H | V4 + (V3 - V4) $\times 2146.5 / 2794.5$ | 37H' | V6 + (V5-V6) × 648.0/2794.5 |
| 38 H | 1 | 1 | 1 | 0 | 0 | 0 | 38H | V4 + (V3 - V4) $\times 2025.0 / 2794.5$ | 38H' | V6 + (V5-V6) $\times 769.5 / 2794.5$ |
| 39H | 1 | 1 | 1 | 0 | 0 | 1 | 39H | V4 + (V3 - V4) × 1863.0/2794.5 | 39H' | V6 + (V5-V6) $\times 931.5 / 2794.5$ |
| 3AH | 1 | 1 | 1 | 0 | 1 | 0 | 3AH | V4 + (V3 - V4) × 1701.0/2794.5 | 3AH' | V6 + (V5-V6) $\times 1093.5 / 2794.5$ |
| 3BH | 1 | 1 | 1 | 0 | 1 | 1 | 3BH | V4 + (V3 - V4) × 1498.5/2794.5 | 3BH' | V6 + (V5-V6) $\times 1296.0 / 2794.5$ |
| 3CH | 1 | 1 | 1 | 1 | 0 | 0 | 3CH | V4 + (V3 - V4) × 1296.0/2794.5 | 3CH' | V6 + (V5-V6) $\times 1498.5 / 2794.5$ |
| 3DH | 1 | 1 | 1 | 1 | 0 | 1 | 3DH | V4 + (V3 - V4) × 1053.0/2794.5 | 3DH' | V7 + (V5-V6) $\times 1741.5 / 2794.5$ |
| 3EH | 1 | 1 | 1 | 1 | 1 | 0 | 3EH | V4 + (V3 - V4) $\times 648.0 / 2794.5$ | 3EH' | V7 + (V5-V6) $\times 2146.5 / 2794.5$ |
| 3FH | 1 | 1 | 1 | 1 | 1 | 1 | 3FH | V4 | 3FH' | V5 |

- Relationship between LOAD and POL output waveforms


| POL | OUT $_{2 N-1}$ | OUT $_{2 N}$ |
| :---: | :---: | :---: |
| L | V0 to V4 | V5 to V9 |
| H | V5 to V9 | V0 to V4 |

(*) $\mathrm{OUT}_{2 \mathrm{~N}-1}$ (odd-numbered outputs) OUT2N (even-numbered outputs)

## Timing Chart

## - Start pulse and data sequence


(*) Upper: OUT1 $\rightarrow$ U/D = High level Lower: OUT379 $\rightarrow$ U/D = Low level

- Loading and cascading operation


 First DATA at

(*) First input

n
Panel

Maximum Ratings $\left(\mathrm{DV}_{\mathrm{SS}}=A V_{S S}=0 \mathrm{~V}\right)$

| Characteristics | Rymbol | Rating | Unit | Applicable Pin |
| :--- | :---: | :---: | :---: | :---: |
| Digital supply voltage | $\mathrm{DV}_{\mathrm{DD}}$ | -0.3 to 4.0 | V |  |
| Analog supply voltage | $\mathrm{AV}_{\mathrm{DD}}$ | -0.3 to 11.0 | V |  |
| Gamma correction voltage | $\mathrm{V}(0: 9)$ | -0.3 to $\mathrm{AV}_{\mathrm{DD}}+0.3$ | V | V 0 to V 9 |
| Digital input voltage | $\mathrm{V}_{\mathbb{I N}}$ | -0.3 to $\mathrm{DV} \mathrm{DD}+0.3$ | V |  |
| Storage temperature | $\mathrm{T}_{\mathrm{Stg}}$ | -55 to 125 | ${ }^{\circ} \mathrm{C}$ |  |

Operating Range $\left(\mathrm{DV}_{\mathrm{SS}}=A \mathrm{~V}_{\mathrm{SS}}=0 \mathrm{~V}\right)$

| Characteristics | Symbol | Rating | Unit | Applicable Pin |
| :---: | :---: | :---: | :---: | :---: |
| Digital supply voltage | DV ${ }_{\text {DD }}$ | 3.0 to 3.6 | V |  |
| Analog supply voltage | AV ${ }_{\text {DD }}$ | 7.5 to 10.0 | V |  |
| Gamma correction voltage | $\mathrm{V}(0: 9)$ | 0.1 to $\mathrm{AV}_{\mathrm{DD}}-0.1$ | V | V0 to V9 |
| Operating temperature | Topr | -20 to 75 | ${ }^{\circ} \mathrm{C}$ |  |
| Operating frequency | $\mathrm{f}_{\mathrm{CPH}}$ | 40 (max) | MHz | CPH |
| Output load capacitance | $\mathrm{C}_{\mathrm{L}}$ | 75 | pF/PIN | OUT1 to OUT384 |
| Input capacitance | $\mathrm{Cl}_{\text {IN }}$ | 8 (max) | pF | DI/O, DO/I |
|  |  | 5 (max) |  | Input pins except DI/O, DO/I |

## Electrical Characteristics

DC Characteristics


| Characteristics |  | Symbol | Test Circuit | Test Condition | Min | Typ. | Max | Unit | Applicable Pin |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input voltage | Low level | $\mathrm{V}_{\text {IL }}$ | - | - | 0 | - | $\begin{aligned} & 0.3 \times \\ & D V_{D D} \end{aligned}$ | V | Logic input |
|  | High level | $\mathrm{V}_{\mathrm{IH}}$ |  | - | $\begin{aligned} & 0.7 \times \\ & D V_{D D} \end{aligned}$ | - | DV ${ }_{\text {DD }}$ |  |  |
| Output voltage | Low level | $\mathrm{V}_{\mathrm{OL}}$ | - | $\mathrm{lOL}=1 \mathrm{~mA}$ | 0 | - | 0.5 | V | Logic output |
|  | High level | $\mathrm{V}_{\mathrm{OH}}$ |  | $\mathrm{l}^{\mathrm{OH}}=-1 \mathrm{~mA}$ | $\begin{gathered} D V_{D D} \\ -0.5 \end{gathered}$ | - | - |  |  |
| Output voltage range |  | VDO | - | - | $\begin{aligned} & A V_{S S} \\ & +0.1 \end{aligned}$ | - | $\begin{gathered} A V_{D D} \\ -0.1 \end{gathered}$ | V | OUT1 to OUT384 |
| Output voltage deviation |  | $\Delta \mathrm{V}_{\mathrm{O}}$ | - | (Note 4) | -20 | - | 20 | mV | OUT1 to OUT384 |
|  |  | (Note 5) |  | -30 | - | 30 |  |  |
| Output amplitude voltage deviation |  |  | $\Delta \mathrm{Vp}-\mathrm{p}$ | - | (Note 6) | -10 | - | 10 | mV | OUT1 to OUT384 |
|  |  | (Note 7) |  |  | -20 | - | 20 | mV |  |  |
|  |  | (Note 8) |  |  | -30 | - | 30 | mV |  |  |
| Gamma resistance fluctuation |  | $\mathrm{R} \gamma$ | - | - | 7.7 | 12.8 | 18 | k $\Omega$ | $\begin{aligned} & \text { V0 to V4 } \\ & \text { V5 to V9 } \end{aligned}$ |  |
| Leakage current |  | IN | - | - | -1 | - | 1 | $\mu \mathrm{A}$ | Logic input |  |
| Standby current |  | IDSTB | - | - | - | - | 1 | $\mu \mathrm{A}$ | DV ${ }_{\text {DD }}$ |  |
| Current dissipation |  | DIDD | - | (Note 9) | - | - | 5 | mA | DV ${ }_{\text {DD }}$ |  |
|  |  | AlDD1 |  |  | - | - | 6 |  | $A V_{D D}$ |  |
|  |  | AldD2 |  |  | - | - | 4 |  |  |  |
|  |  | AldD3 |  | (Note 10) | - | - | 16 |  |  |  |
|  |  | $\mathrm{Al}_{\text {DD4 }}$ |  |  | - | - | 16 |  |  |  |
| Output current |  | I chg | - | $\begin{aligned} & \text { Vout }=8.9 \mathrm{~V}, \mathrm{AV}_{\mathrm{DD}}=9 \mathrm{~V} \\ & \mathrm{Vx}=4 \mathrm{~V} \quad \text { (Note 11) } \end{aligned}$ | - | - | -20 | $\mu \mathrm{A}$ | OUT1 to OUT384 |  |
|  |  | $I_{\text {dis }}$ | - | $\begin{aligned} & \text { Vout }=0.1 \mathrm{~V}, \mathrm{AV}_{\mathrm{DD}}=9 \mathrm{~V} \\ & \mathrm{Vx}=1 \mathrm{~V} \quad \\ & \text { (Note 11) } \end{aligned}$ | 30 | - | - |  |  |  |

Note4: $\quad A V_{D D}=9 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}+1.5 \mathrm{~V} \leq \mathrm{VOUT} \leq \mathrm{AV}_{\mathrm{DD}}-1.5 \mathrm{~V}$
$\Delta \mathrm{VO}$ is the numerical different the anticipated value of LCD panel drive output voltage (refer for (7) relationship between grayscale data and output voltage) from each LCD panel drive output voltage, These relationship shows as following formula.
$\Delta \mathrm{VO}=$ [each LCD panel drive pin output voltage] - [anticipated value of LCD panel drive output voltage]
Note 5: $A V_{D D}=9 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}+0.1 \mathrm{~V} \leq \mathrm{VOUT}<\mathrm{AV}_{\mathrm{SS}}+1.5 \mathrm{~V}, ~ \mathrm{AV}_{\mathrm{DD}}-1.5 \mathrm{~V}<\mathrm{VOUT} \leq \mathrm{AV}_{\mathrm{DD}}-0.1 \mathrm{~V}$ The formula for $\Delta \mathrm{VO}$ is same as Note 4.

Note 6: $A V_{D D}=9 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}+1.5 \mathrm{~V}<\mathrm{VOUT}<\mathrm{AV}_{\mathrm{DD}}-1.5 \mathrm{~V}$
$\Delta \mathrm{Vpp}$ is the numerical different the remainder of the average of all LCD panel drive output voltage at positive electrode and negative electrode from each LCD panel drive output voltage
$\Delta \mathrm{Vpp}=\{[$ the remainder of each LCD panel drive output voltage at positive electrode ( V 0 to V 4 ) and negative electrode (V5 to V9)]- [all LCD panel drive output voltage at positive electrode (V0 to V4) and negative electrode (V5 to V9) of LCD panel drive output voltage]\} in the same grayscale

Note 7: $\mathrm{AV}_{\mathrm{DD}}=9 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}+0.8 \mathrm{~V}<\mathrm{VOUT} \leq \mathrm{AV}_{\mathrm{SS}}+1.5 \mathrm{~V}, \mathrm{AV}_{\mathrm{DD}}-1.5 \mathrm{~V} \leq \mathrm{VOUT}<\mathrm{AV}_{\mathrm{DD}}-0.8 \mathrm{~V}$ The formula for $\Delta \mathrm{Vpp}$ is same as Note 6.

Note 8: $A V_{D D}=9 \mathrm{~V}, \mathrm{AV}_{S S}=0 \mathrm{~V}, \mathrm{AV}_{\mathrm{SS}}+0.1 \mathrm{~V} \leq \mathrm{VOUT} \leq \mathrm{AV}_{\mathrm{SS}}+0.8 \mathrm{~V}, \mathrm{AV} \mathrm{VD}_{\mathrm{DD}}-0.8 \mathrm{~V} \leq \mathrm{VOUT} \leq \mathrm{AV}_{\mathrm{DD}}-0.1 \mathrm{~V}$ The formula for $\Delta \mathrm{Vpp}$ is same as Note 6.

Note 9: LOAD cycle $=20 \mu \mathrm{~s}, \mathrm{f} \mathrm{CPH}=32.5 \mathrm{MHz}$
DIDD: dot-checkered input pattern, no loaded
AlDD1: V0 to $\mathrm{V} 4=\mathrm{V} 5$ to $\mathrm{V} 9=4.5 \mathrm{~V}, \mathrm{AV} D=9 \mathrm{~V}$, LCD panel drive load $=200 \Omega+80 \mathrm{pF}$, $\overline{\mathrm{TESTP}}=$ " H "
AldD2: V 0 to $\mathrm{V} 4=\mathrm{V} 5$ to $\mathrm{V} 9=4.5 \mathrm{~V}, \mathrm{AV}_{\mathrm{DD}}=9 \mathrm{~V}$, LCD panel drive load $=200 \Omega+80 \mathrm{pF}$, $\overline{\mathrm{TESTP}}=$ " L "
Note 10: LOAD cycle $=20 \mu \mathrm{~s}, \mathrm{f} C P \mathrm{H}=32.5 \mathrm{MHz}$
AldD3: $\mathrm{V} 0=8.9 \mathrm{~V}, \mathrm{~V} 4=\mathrm{V} 5, \mathrm{~V} 9=0.1 \mathrm{~V}, \mathrm{AV}$ DD $=9 \mathrm{~V}$, LCD panel drive load $=200 \Omega+80 \mathrm{pF}$, $\overline{\mathrm{TESTP}}=$ "H"
$A_{D D} 4: V 0=8.9 \mathrm{~V}, \mathrm{~V} 4=\mathrm{V} 5, \mathrm{~V} 9=0.1 \mathrm{~V}, \mathrm{AV}_{\mathrm{DD}}=9 \mathrm{~V}, \mathrm{LCD}$ panel drive load $=200 \Omega+80 \mathrm{pF}$, $\overline{\mathrm{TESTP}}=$ " L "
Note 11: Voltage applied to LCD panel drive



AC Characteristics
$\left(D V_{D D}=3.0\right.$ to $3.6 \mathrm{~V}, \mathrm{AV}$ DD $=7.5$ to $10.0 \mathrm{~V}, \mathrm{DV}_{\mathrm{SS}}=\mathrm{AV} \mathrm{SS}=0 \mathrm{~V}, \mathrm{Ta}=-20$ to $\left.75^{\circ} \mathrm{C}\right)$

| Characteristics | Symbol | Test Circuit | Test Condition | Min | Typ. | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CPH pulse width H | tcWh | - | - | 4 | - | - | ns |
| CPH pulse width L | tcWL | - | - | 4 | - | - | ns |
| Enable setup time | $\mathrm{t}_{\text {SDI }}$ | - | - | 4 | - | - | ns |
| Enable hold time | thDI | - | - | 0 | - | - | ns |
| Data DINV setup time | $\mathrm{t}_{\text {sDD }}$ | - | - | 4 | - | - | ns |
| Data DINV hold time | thDD | - | - | 0 | - | - | ns |
| LOAD high period | tLWH | - | - | 3 | - | - | CPH <br> cycle |
| LOAD enable input period | tLOAD-D1 | - | - | 2 | - | - | CPH <br> cycle |
| LOAD enable output period | tLD-LOAD | - | - | 1 | - | - | $\mathrm{CPH}$ <br> cycle |
| LOAD setup time | $t_{\text {sLD }}$ | - | - | 6 | - | - | ns |
| POL setup time | $t_{\text {sDP }}$ | - | - | 4 | - | - | ns |
| POL hold time | thDP | - | - | 6 | - | - | ns |
| Enable output delay time | $t_{\text {pdDO }}$ | - | $C_{L}=15 \mathrm{pF}$ | - | - | 15 | ns |
| Output delay time 1 | $t_{\text {pdDE }}$ | - | $\begin{aligned} & \mathrm{C}_{\mathrm{L}}=75 \mathrm{pF} \\ & \mathrm{R}=5 \mathrm{k} \Omega \\ & \text { Target output voltage } \times 0.9 \\ & \quad(\text { Note } 12) \end{aligned}$ | - | - | 6 | $\mu \mathrm{S}$ |
| Output delay time 2 | $t_{\text {pdD }}$ | - | $\begin{aligned} & C_{L}=75 \mathrm{pF} \\ & \mathrm{R}=5 \mathrm{k} \Omega \\ & \text { Target output voltage } \pm \Delta \mathrm{V}_{\mathrm{O}} \\ & \quad(\text { Note } 12) \end{aligned}$ | - | - | 11 | $\mu \mathrm{S}$ |

Note 12: Output load condition



Note: Timing for loading OUT379 to OUT384
Timing for loading OUT1 to OUT6



## Power-On Sequence

At power ON, the sequence is: $\mathrm{DVDD} \rightarrow$ LOGOC input signal $\rightarrow \mathrm{AVDD}$, reference analog voltage At power OFF, the sequence is the reverse of that for power ON.
Turn ON/OFF power supplies for AV DD and reference analog voltage inputs simultaneously.
As long as the voltage condition, $\mathrm{AV} D \mathrm{D}>\mathrm{DVDD}$, is satisfied, all power supplies can be turned OFF simultaneously.


## RESTRICTIONS ON PRODUCT USE

- TOSHIBA is continually working to improve the quality and reliability of its products. Nevertheless, semiconductor devices in general can malfunction or fail due to their inherent electrical sensitivity and vulnerability to physical stress. It is the responsibility of the buyer, when utilizing TOSHIBA products, to comply with the standards of safety in making a safe design for the entire system, and to avoid situations in which a malfunction or failure of such TOSHIBA products could cause loss of human life, bodily injury or damage to property.
In developing your designs, please ensure that TOSHIBA products are used within specified operating ranges as set forth in the most recent TOSHIBA products specifications. Also, please keep in mind the precautions and conditions set forth in the "Handling Guide for Semiconductor Devices," or "TOSHIBA Semiconductor Reliability Handbook" etc..
- The TOSHIBA products listed in this document are intended for usage in general electronics applications (computer, personal equipment, office equipment, measuring equipment, industrial robotics, domestic appliances, etc.). These TOSHIBA products are neither intended nor warranted for usage in equipment that requires extraordinarily high quality and/or reliability or a malfunction or failure of which may cause loss of human life or bodily injury ("Unintended Usage"). Unintended Usage include atomic energy control instruments, airplane or spaceship instruments, transportation instruments, traffic signal instruments, combustion control instruments, medical instruments, all types of safety devices, etc.. Unintended Usage of TOSHIBA products listed in this document shall be made at the customer's own risk.
- Polyimide base film is hard and thin. Be careful not to injure yourself on the film or to scratch any other parts with the film. Try to design and manufacture products so that there is no chance of users touching the film after assembly, or if they do, that there is no chance of them injuring themselves. When cutting out the film, try to ensure that the film shavings do not cause accidents. After use, treat the leftover film and reel spacers as industrial waste
- Light striking a semiconductor device generates electromotive force due to photoelectric effects. In some cases this can cause the device to malfunction.
This is especially true for devices in which the surface (back), or side of the chip is exposed. When designing circuits, make sure that devices are protected against incident light from external sources. Exposure to light both during regular operation and during inspection must be taken into account.
- The products described in this document are subject to the foreign exchange and foreign trade laws.
- The information contained herein is presented only as a guide for the applications of our products. No responsibility is assumed by TOSHIBA CORPORATION for any infringements of intellectual property or other rights of the third parties which may result from its use. No license is granted by implication or otherwise under any intellectual property or other rights of TOSHIBA CORPORATION or others.
- The information contained herein is subject to change without notice.

