# **MOTOROLA** SEMICONDUCTOR **TECHNICAL DATA**

Product Preview

# **Automotive Dual High** Side TMOS Driver

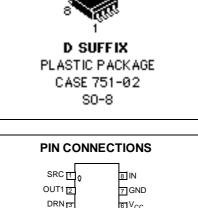
The MC33285 is a dual high side TMOS driver designed for use in the harsh automotive switching applications.

The purpose of the MC33285 is to drive two power n-channel FETs in a high side application with an inductive load. The application in an automotive environment requires the capability of withstanding high voltages and load dump transients. The MC33285 is able to withstand reverse battery conditions at selected pins.

It protects the n-channel power FET on OUT1 under over current condition. This device has one input to control both stages on or off.

- Temperature Range from -40°C to +125°C
- PWM Capability
- Power TMOS # 1 Over-Current and Short-Circuit Protection
- Voltage Range 7V to 40V
- Extended Temperature Range from -40°C to 125°C
- Load Dump Protected
- Over Voltage Detection and Activation of OUT2 during Overvoltage
- Single Input Control for both Output Stages
- Capacitor Value of 100nF Connected to Pin CP
- Analog Input Control Measurement Detection
- OUT1 LOAD leakage measurement detection

right to change or discontinue this product without notice



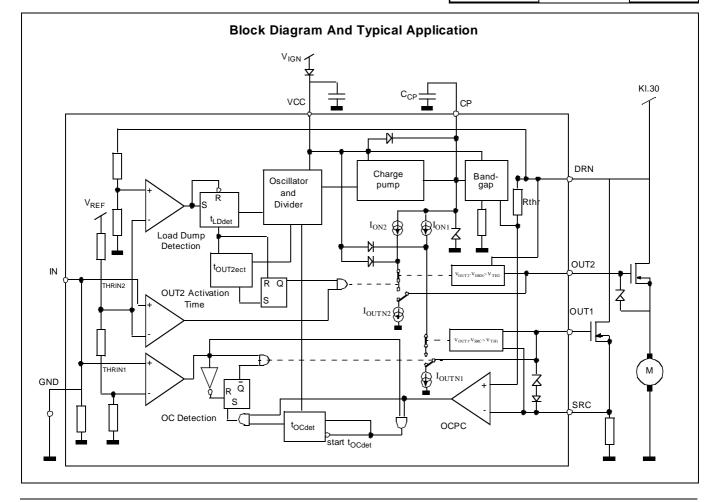
# **ORDERING INFORMATION**

OUT2

6 V<sub>CC</sub>

5 CP

| Device   | Temperature<br>Range | Package |
|----------|----------------------|---------|
| MC33285D | -40°C to +125°C      | SO8     |





**HIGH SIDE** TMOS DRIVER

SILICON MONOLITHIC INTEGRATED CIRCUIT

# MC33285

#### MC33285

## MAXIMUM RATING

Operating ambient temperature

| Rating   | Symbol           | Value                 | Unit |
|--|------------------|-----------------------|------|
| Maximum voltage at pins OUT1 OUT2                          | V <sub>OUT</sub> | V <sub>VCC</sub> + 20 | V    |
| Maximum voltage at pin CP                                  | V <sub>CP</sub>  | 50                    | V    |
| Input voltage V <sub>i</sub> at DRN                        | V <sub>DRN</sub> | -2 ; 40               | V    |
| Input voltage V <sub>i</sub> at SRC                        | V <sub>SRC</sub> | -5 ; 40               | V    |
| Input voltage at pin $V_{CC}$                              | V <sub>CC</sub>  | -2 to 40              | V    |
| Input Voltage at pin IN. Condition : $-2V < V_{VCC} < 40V$ | V <sub>IN</sub>  | -2 to $V_{VCC}$       | V    |
| Operational voltage $V_{VCC}$ at pin $V_{CC}$              | V <sub>VCC</sub> | 7 to V <sub>i</sub>   | V    |
| THERMAL RATINGS  |                  |                       |      |
| Rating   | Symbol           | Value                 | Unit |
| Storage Temperature  | T <sub>stg</sub> | -40 to +150           | °C   |

 $\label{eq:Electrical characteristics.} (\underline{T_A \ from \ -40^\circ C \ to \ +125^\circ C}, \ Vcc \ from \ 7V \ to \ 20V, \ unless \ otherwise \ noted)$ 

 $\mathsf{T}_{\mathsf{a}}$ 

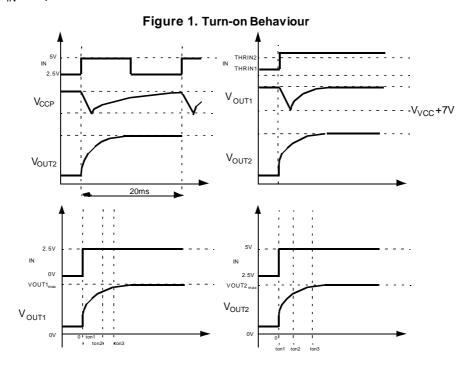
°C

-40 to +125

| Characteristic   | Symbol                | Min  | Тур | Max        | Unit |
|--|-----------------------|------|-----|------------|------|
| OVERVOLTAGE AND OVERCURRENT  |                       |      |     |            |      |
| Load Dump Detection Time   | tLD <sub>det</sub>    | 250  | 400 | 550        | μs   |
| Load Dump Activation Time  | tOUT2act              | 300  | 460 | 620        | ms   |
| Error Voltage Threshold  | VDRN - VSRC           | 1.12 |     | 1.44       | V    |
| Overcurrent Detection Time   | tOCdet                | 520  | 800 | 1080       | μs   |
| SRC PIN 1  |                       |      |     |            |      |
| Leakage Current  | ILC <sub>det</sub>    | 15   | 30  | 50         | mA   |
| Leakage Current Detection Time   | tLC <sub>det</sub>    | 130  | 200 | 270        | μs   |
| DRN PIN 3  |                       |      |     |            |      |
| Operating Current (7V <v<sub>DRN&lt;20V)</v<sub>   | I <sub>DRN</sub>      |      |     | 1.5        | mA   |
| Leakage current (0V <v<sub>DRN&lt;20V, V<sub>VCC</sub>&lt;4V)</v<sub>  | I <sub>leak_DRN</sub> | -5   |     | 5          | μA   |
| OUT1 PIN2, OUT 2 PIN 4   |                       |      |     |            |      |
| Output On Voltage. Charge Pump ON  | Von                   |      |     | Vcc+15     | V    |
| Turn off current, Vout >0.5V   | loutoff               | 66   | 110 | 154        | μΑ   |
| Turn On Time, OUT1: 8nF,10μA; OUT2:<br>16nF,10μA<br>- 7V < Vcc < 10V, Vout > Vcc+7<br>- 10V < Vcc < 20V, Vout > Vcc+11 | ton                   |      |     | 1.5<br>1.5 | ms   |
| VCC PIN 6  |                       |      |     |            |      |
| Supply Voltage Range   | Vcc                   | 7    |     | 40         | V    |
| Quiescent Supply Current at Vcc = 20V  | lcc                   |      |     | 10         | mA   |
| IN PIN 8   |                       | ·    | I   |            |      |
| Input Low Voltage OUT1   | Vil                   |      |     | 0.7        | V    |
| Input High Voltage OUT1  | Vih                   | 1.7  |     |            | V    |
| Input Hysteresis OUT1, OUT2  | Vhys                  | 0.4  |     |            | V    |
| Input Pull Down Current, 0.7V <vin<6v< td=""><td>lin</td><td>7.5</td><td>15</td><td>16.5</td><td>μΑ</td></vin<6v<>     | lin                   | 7.5  | 15  | 16.5       | μΑ   |
| Open Input Voltage   | Viop                  |      |     | 0.7        | V    |
| Input Low Voltage OUT2   | Vil2                  |      |     | 3          | V    |
| Input High Voltage OUT2  | Vih2                  | 3.9  |     |            | V    |

### **Turn On Characteristics**

The power FETs are turned on by charging their gate capacities with a current flowing out of pin OUT1 and OUT2. During PWM, the values of table below are guaranteed. They are measured with 8nF on OUT1 and 16nF on OUT2. - test condition :  $V_{IN}$  : ramp 0V to 2.5V or 2.5V to 5V.



#### Table 1 Turn on behaviour

| Voltage V <sub>VCC</sub>   | Minimum V <sub>OUT1,OUT2</sub>  | Minimum V <sub>OUT1,OUT2</sub>                       | Minimum V <sub>OUT1,OUT2</sub>                  |
|--|---|--|---|
|  | after t <sub>on1</sub> = 100µsec  | after t <sub>on2</sub> = 1.0msec                     | after t <sub>on3</sub> = 1.5msec                |
| $7V < V_{VCC} < 10V$<br>10V < $V_{VCC} < 20V$<br>20V < $V_{VCC} < 40V$ | V <sub>VCC</sub> - 0.7V<br>V <sub>VCC</sub> - 0.7V<br>V <sub>VCC</sub> - 0.7V | V <sub>VCC</sub> + 5.95V<br>V <sub>VCC</sub> + 9.35V | V <sub>VCC</sub> + 7V<br>V <sub>VCC</sub> + 11V |

The output voltages at OUT1 and OUT2 are limited by controlling the current sources  $I_{on1}$ ,  $I_{on2}$  to avoid current flowing through the external or the internal zener diode.

If Vcc + Vth (threshold voltage) is reached, the current sources are turned off.

- threshold V  $_{TH1}$  for OUT1 output voltage control : 7V < V  $_{TH1}$  < Vz

- threshold VTH2 for OUT2 output voltage control :  $7V < V_{TH2} < 15V$ 

#### **Turn off characteristics**

The power FETs on OUT1 and OUT2 are turned off by discharging the gate capacity with the constant discharge current  $I_{OUToff}$ .

- discharge current  $I_{OUTxoff}$  : IOUTxoff = 110µA

condition :  $V_{OUT}x > 0.5V (V_{IN} < V_{THRxIN})$ 

Test conditions for switching off the power FETs :

- 1. IN open

- 2. Stages disabled via pin IN
- 3. Stage OUT1 disabled by an over current error

### FUNCTIONNAL DESCRIPTION

#### Introduction

The MC33285 contains only one charge pump for two outputs. The outputs OUT1 and OUT2 are switched on and off by the input IN .There are three ways to control the outputs: OUt1 can be switched alone , they can be switched together or OUT2 can be switched when OUT1 is already on .In the last case , the voltage drop on OUT1 when charging OUT2 is limited .

The external capacitor  $C_{CP}$  connected to pin CP is used to store the charge continuously delivered by the charge pump . The voltage on this pin is limited to a maximum value  $V_{CPmax}$ . Both outputs are sourced with a constant current from  $C_{CP}$  to switch them on . In addition , the gates of the power FETs are precharged from VCC to prevent  $C_{CP}$  from being discharged by a voltage on OUT1 or OUT2 which is still lower than  $V_{VCC}$ . The values of the output voltages are limited to  $V_{OUT1max}$  and  $V_{OUT2max}$ 

The power FET on OUT1 is protected againts an exceeded gate-source voltage by an internal zener diode.

Channel 1 allows to protect the n-channel power FET on OUT1 under over current condition. The drain-source voltage of the FET on OUT1 will be checked, if the channel 1 is switched on. The internal error voltage threshold determines the maximum drain-source voltage that allows the power FET to stay in the on state. If the measured drain-source voltage exceeds the internal error voltage threshold, the output of the Over Current Protection Comparator (OCPC) is enabled. If the output of the OCPC is active longer than  $t_{OCdet}$ , the output OUT1 is switched off.

After switching off the power FET on OUT1 by an over current condition, the power FET can only be turned on again by the input IN.

When switching off the power FETs their gate capacities are discharged by a constant currennt  $I_{OUToff}$ 

IF the input IN is disconnected, the MC33285 outputs OUT1 and OUT2 are in the off state.

If overvoltage occurs on pin DRN for a time period longer than  $t_{LDdet}$ , then OUT2 is switched on for the time  $t_{OUT2act}$ . In overvoltage condition OUT1 is off if IN is below Vih.

#### Internal Zener Diode

An on-chip zener diode is placed between OUT1 and SRC. Design guarantees that  $V_Z\!>\!V_{TH1}$ 

- zener clamping voltage between OUT1 and SRC :  $V_{TH1}$  <  $V_Z$  < 20V

#### **PWM** capability

The CPIC2 is PWM capable on OUT2. The loss of charge on Ccp when switching on OUT2 is refreshed until the start on the next PWM cycle to a value which is sufficient to guarantee the specified turn on behaviour.

The PWM capability is measured with a test circuit and load conditions  $% \label{eq:pwm} \left( \mathcal{A}_{i}^{(i)} \right) = \left( \mathcal{A}_{i}^{(i)} \right) \left$ 

- PWM cycle : period T = 20ms ; OUT2 is switched on from 10% to 90% of T .

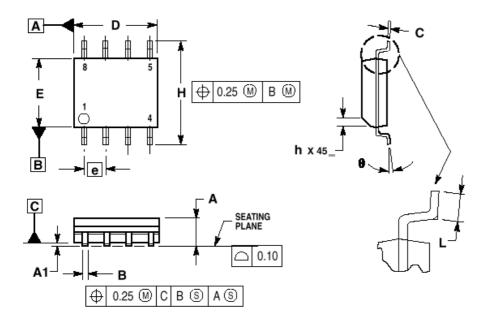
- Test condition :  $V_{\mbox{\scriptsize IN}}$  : ramps 2.5V to 5V according to PWM cycle defined above.

## Crosstalk between OUT1 and OUT2

If output OUT2 is switched on while OUT1 is already on, the voltage drop that occurs on OUT1 is limited.

Voltage drop on OUT1 :  $10V < V_{VCC} < 20V$  : OUT1 not below  $V_{VCC} + 7V$  $7V < V_{VCC} < 20V$  : OUT1 not below  $V_{VCC} + 7V$ 

Each time OUT1 is switched on, a current ILCdet is sourced out of pin SRC for the time tLCdet to check if there is an external leakage current on that node in the application. The high side switch on OUT1 is turned on only if the test is successful."



NOTES:

- DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
- 2. DIMENSIONS ARE IN MILLIMETER.
- DIMENSION D AND E DO NOT INCLUDE MOLD PROTRUSION.
- MAXIMUM MOLD PROTRUSION 0.15 PER SIDE.
  DIMENSION B DOES NOT INCLUDE DAMBAR

PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 TOTAL IN EXCESS OF THE B DIMENSION AT MAXIMUM MATERIAL CONDITION.

|     | MILLIMETERS |      |  |
|-----|-------------|------|--|
| DIM | MIN         | MAX  |  |
| Α   | 1.35        | 1.75 |  |
| A1  | 0.10        | 0.25 |  |
| В   | 0.35        | 0.49 |  |
| С   | 0.19        | 0.25 |  |
| D   | 4.80        | 5.00 |  |
| Е   | 3.80        | 4.00 |  |
| e   | 1.27 BSC    |      |  |
| н   | 5.80        | 6.20 |  |
| h   | 0.25        | 0.50 |  |
| L   | 0.40        | 1.25 |  |
| 8   | 0_          | 7_   |  |

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