

# General Description

The MAX850–MAX853 low-noise, inverting, chargepump power supplies are ideal for biasing GaAs FETs in cellular telephone transmitter amplifiers.

The MAX850–MAX852 offer both preset (-4.1V) and adjustable (-0.5V to -9.0V) output voltages. The MAX853 uses an external positive control voltage to set the negative output voltage. Input voltage range for all four devices is 4.5V to 10V. Output current is 5mA.

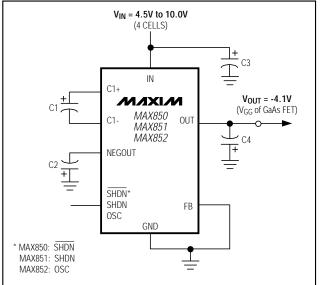
An internal linear regulator reduces the output voltage ripple to 2mVp-p. With a well-filtered control voltage (VCTRL), the MAX853 achieves typical output ripple of less than 1mVp-p. Supply current is 3mA max and shutdown current is less than  $1\mu A$  max over temperature ( $5\mu A$  max for MAX851).

# **Applications**

Cellular Phones

- Negative Regulated Power Supplies Personal Communicators, PDAs
- Wireless Data Loggers
- Continuously Adjustable GaAs FET Bias
- LCD-Bias Contrast Control





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## Features

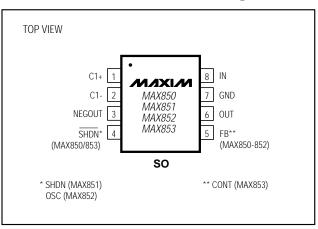
- + Fixed -4.1V or Adjustable -0.5V to -9V Output at 5mA
- 4.5V to 10V Input Voltage Range
- 2mVp-p Output Voltage Ripple (MAX850–MAX852) 1mVp-p Output Voltage Ripple (MAX853)
- 100kHz Charge-Pump Switching Frequency (MAX850/MAX851/MAX853)
- External Synchronizing Clock Input (MAX852)
- Logic-Level Shutdown Mode: 1µA Max Over Temperature (MAX850/MAX852/MAX853)
- Low Cost, 8-Pin SO Package

# **Ordering Information**

PART	TEMP. RANGE	PIN-PACKAGE			
MAX850CSA	0°C to +70°C	8 SO			
MAX850C/D	0°C to +70°C	Dice*			
MAX850ESA	-40°C to +85°C	8 SO			
MAX851CSA	0°C to +70°C	8 SO			
MAX851C/D	0°C to +70°C	Dice*			
MAX851ESA	-40°C to +85°C	8 SO			
MAX852CSA	0°C to +70°C	8 SO			
MAX852C/D	0°C to +70°C	Dice*			
MAX852ESA	-40°C to +85°C	8 SO			
MAX853CSA	0°C to +70°C	8 SO			
MAX853C/D	0°C to +70°C	Dice*			
MAX853ESA	-40°C to +85°C	8 SO			
*Dice are specified at $T_A = +25^{\circ}C$ only.					

PART	TEMP. RANGE	BOARD TYPE		
MAX850EVKIT-SO	0°C to +70°C	Surface Mount		

# \_Pin Configuration



Maxim Integrated Products 1

## **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage, VIN to GND	0.3V to 10.5V
VNEGOUT to GND	10.5V to 0.3V
VIN to VNEGOUT	0.3V to 21V
VOUT to GND (Note 1)	
SHDN or OSC (pin 4) Voltage to GND	0.3V to (V <sub>IN</sub> + 0.3V)
Continuous Power Dissipation ( $T_A = +70^{\circ}$ C	C)
SO (derate 5.88mW/°C above +70°C)	471mW

Operating Temperature Ranges	
MAX85_C	0°C to +70°C
MAX85_ESA	40°C to +85°C
Storage Temperature Range	65°C to +160°C
Lead Temperature (soldering, 10sec)	+300°C

Note 1: The output may be shorted to NEGOUT or GND if the package power dissipation is not exceeded. Typical short-circuit current to GND is 50mA.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

# ELECTRICAL CHARACTERISTICS

 $(+5V \le V_{IN} \le +10V, \text{GND} = 0V, V_{OUT} = -4.1V, \text{R}_{L} = \infty, \text{T}_{A} = \text{T}_{MIN}$  to T<sub>MAX</sub>, unless otherwise noted. A 100kHz, 50% duty cycle square wave between GND and VIN is applied to the OSC pin of the MAX852.)

PARAMETER	SYMBOL CONDITIONS		MIN	TYP	MAX	UNITS	
Supply Voltage Range (Note 2)	VIN		5		10	V	
Output Voltage	Vout	MAX850–MAX852, $V_{FB} = 0V$ , $R_L = \infty$ or 820 $\Omega$ , Figure 2a	-4.3	-4.1	-3.9	V	
	VOUT	MAX853, $V_{CTRL} = 4.1V$ , R <sub>L</sub> = $\infty$ or 820 $\Omega$ , Figure 2c	-4.2		-4.0		
Output Voltage Range			-0.5 to - (V <sub>IN</sub> - 1)		V		
Set Voltage	V <sub>FBset</sub>	MAX850–MAX852, no load, Figure 2b	-1.32	-1.28	-1.24	V	
Supply Current	ΙQ			2.0	3.0	mA	
		MAX850/MAX853, $V_{IN} = 10V$ , $\overline{SHDN} = 0V$		0.002	1		
Shutdown Supply Current	ISHUT	MAX851, SHDN = 2V		2	5	5 μA 1	
		MAX852, OSC low			1		
		MAX850–MAX852, $V_{FB} = 0V$ , R <sub>L</sub> = $\infty$ or 820 $\Omega$ , Figure 2a		4	8		
V <sub>OUT</sub> Load Regulation		MAX853, $V_{CTRL} = 4.1V$ , R <sub>L</sub> = $\infty$ or 820 $\Omega$ , Figure 2c		3	8	— mV/mA	
V <sub>OUT</sub> Ripple		MAX850-MAX852	2 1				
		MAX853			mVp-p		
Oscillator Frequency (Note 3)	fosc	MAX850/MAX851/MAX853, T <sub>A</sub> = +25°C	80	100	120	kHz	
Input High Voltage	Vih	Pin 4	2.0			V	
Input Low Voltage	VIL	Pin 4			0.5	V	
Input Current	lin	Pin 4			±1	μA	
Input Capacitance	CIN	Pin 4		10		pF	

Note 2: The supply voltage can drop to 4.5V, but the output may no longer sink 5mA at -4.1V.

Note 3: The MAX852 will operate with a 50kHz to 250kHz square wave of 40% to 60% duty cycle. For best performance, use an 80kHz to 120kHz square wave with 50% duty cycle.

**Typical Operating Characteristics** 

(Circuit of Figure 2a,  $V_{IN} = 6V$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

**OUTPUT VOLTAGE vs. OUTPUT VOLTAGE** MAXIMUM OUTPUT CURRENT INPUT VOLTAGE OVER TEMPERATURE vs. OUTPUT CURRENT vs. INPUT VOLTAGE -4.13 60 -4.16  $I_{OUT} = 2.5 \text{mA}$ -4.12 (mA) -4.15 50 () -4.14 -4.13 -4.12 -4.12 MAXIMUM OUTPUT CURRENT €-4.11  $T_A = +25^{\circ}C$ -4.10 -4.09 -4.08 40 -40°C Τ<sub>A</sub> 30  $T_A = +85^{\circ}C$ 20 -4.11 -4.07 10 -4.06 -4.10 9.0 10.0 5.0 6.0 7.0 8.0 0 2.0 4.0 6.0 8.0 10.0 7.0 10.0 5.0 6.0 8.0 9.0 INPUT VOLTAGE (V) OUTPUT CURRENT (mA) INPUT VOLTAGE (V) **NO-LOAD SUPPLY CURRENT** SUPPLY CURRENT START-UP TIME vs. vs. TEMPERATURE vs. INPUT VOLTAGE **INPUT VOLTAGE** 2.00 2.00 2.5 V<sub>IN</sub> = 10.0V I<sub>OUT</sub> = 5mA 2.4 1.80 (W) 1.80 1.60 1.40 1.40 1.20 1.00 0.80 1.60 2.3 SUPPLY CURRENT (mA) ີ ຍິ 1.40 2.2 Ŭ ₩ 1.20 2.1 1.00 1.00 0.80 2.0 1.9 1.8 0.60 1.7 0.40 0.60 0.20 1.6 -20 5.0 6.0 7.0 8.0 9.0 10.0 -40 0 20 40 60 80 100 5.0 6.0 7.0 8.0 9.0 10.0 TEMPERATURE (°C) INPUT VOLTAGE (V) INPUT VOLTAGE (V) **EFFICIENCY vs. LOAD CURRENT** 80 V<sub>IN</sub> = 5.0V 70  $V_{IN} = 6.0V$ 60 EFFICIENCY (%) 50 40  $V_{IN} = 10.0V$ 30 20 10 1.0 1.5 2.0 2.5 3.0 3.5 4.0 4.5 5.0 LOAD CURRENT (mA)

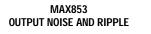
# MAX850-MAX853

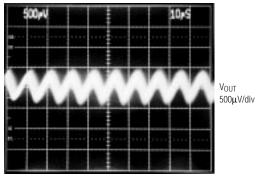
## \_Typical Operating Characteristics (continued)

(Circuit of Figure 2a,  $V_{IN} = 6V$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

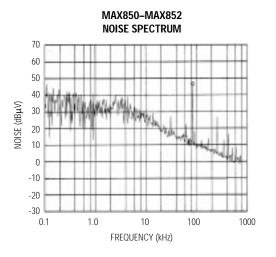
MAX850–MAX852 OUTPUT NOISE AND RIPPLE

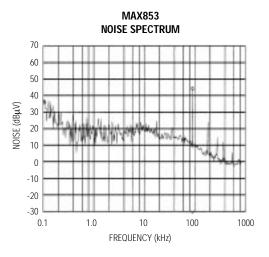
 $10 \mu s/div \label{eq:VIN} V_{IN} = 6.0V, \, V_{OUT} = -4.1V, \, I_{OUT} = 5mA, \, AC \, COUPLED$ 

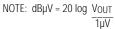




 $10 \mu s/div \label{eq:VIN} V_{IN}$  = 6.0V,  $V_{OUT}$  = -4.1V,  $I_{OUT}$  = 5mA, AC COUPLED





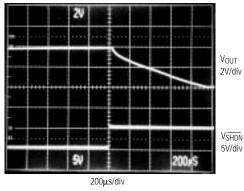


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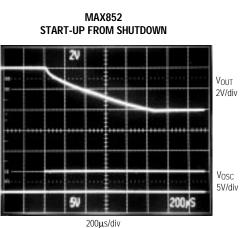
## Typical Operating Characteristics (continued)

(Circuit of Figure 2a,  $V_{IN} = 6V$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

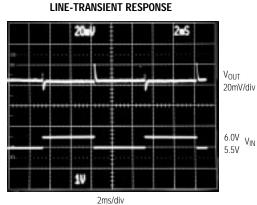
#### MAX850/MAX851/MAX853 START-UP FROM SHUTDOWN



CIRCUIT OF FIGURE 2a,  $V_{IN}$  = 6.0V,  $V_{OUT}$  = -4.1V,  $I_{OUT}$  = 5mA

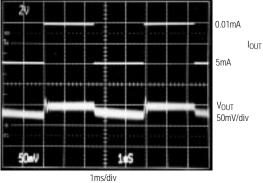


CIRCUIT OF FIGURE 2a,  $V_{IN}$  = 6.0V,  $V_{OUT}$  = -4.1V,  $I_{OUT}$  = 5mA SHUTDOWN OCCURS WHEN 100kHz EXTERNAL CLOCK IS GATED OFF



V<sub>OUT</sub> = -4.08V, I<sub>OUT</sub> = 5mA, AC COUPLED

#### LOAD-TRANSIENT RESPONSE



1n V<sub>OUT</sub> = -4.0V, AC COUPLED

## \_Pin Description

PIN				NAME	FUNCTION
MAX850	MAX851	MAX852	MAX853	NAWE	FUNCTION
1	1	1	1	C1+	Positive terminal for C1
2	2	2	2	C1-	Negative terminal for C1
3	3	3	3	NEGOUT	Negative Output Voltage (unregulated), V <sub>NEGOUT</sub> = -V <sub>IN</sub> + 0.2V
4	-	-	4	SHDN	Active-low TTL logic level Shutdown Input
_	4	-	-	SHDN	Active-high TTL logic level Shutdown Input
-	-	4	-	OSC	External Clock Input
5	5	5	-	FB	Dual-Mode Feedback Input. When FB is grounded, the output is preset to -4.1V. To select other output voltages, connect FB to an external resistor divider. See Figure 2b.
-	-	-	5	CONT	Control Voltage Input. To set V <sub>OUT</sub> , connect a resistor divider between OUT and a positive control voltage between 0V and 10V. See Figure 2c.
6	6	6	6	OUT	Output Voltage
7	7	7	7	GND	Ground
8	8	8	8	IN	Positive Power-Supply Input (4.5V to 10V)

## **Detailed Description**

The MAX850–MAX853 are low-noise, inverting, regulated charge-pump power supplies designed for biasing GaAs FET devices, such as power-amplifier modules in cellular handsets.

The applied input voltage ( $V_{IN}$ ) is first inverted to a negative voltage at NEGOUT by a capacitive charge pump. This voltage is then regulated by an internal linear regu

C1-+ CHARGE PUMP MAXIM C1 MAX850 MAX851 MAX852 NEGOUT Ν J Ŧ + SHDN (MAX850) SHDN (MAX851) 0 CONNECT TO OSC (MAX852) GND TO SET  $V_{OUT} = -4.1V$ -1.28V REF GND

Figure 1a. MAX850–MAX852 Block Diagram

lator, and appears at OUT (Figure 1). The minimum (most negative) output voltage (V<sub>OUT</sub>) achievable is the inverted positive voltage, plus the 1.0V required by the post-regulator. The ripple noise induced by the charge-pump inverter is reduced by the linear regulator to 2mVp-p at V<sub>OUT</sub> for the MAX850–MAX852. In addition, the excellent AC rejection of the linear regulator attenuates noise on the incoming supply. Up to 5mA is available at OUT.

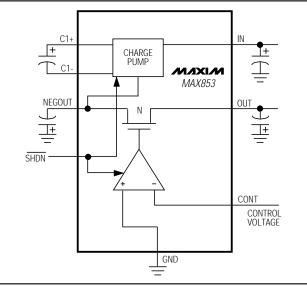


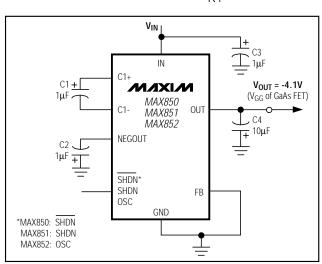
Figure 1b. MAX853 Block Diagram

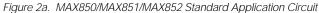


# Applications Information Setting the Output Voltage

For the MAX850–MAX852, select either a fixed or an adjustable output voltage. Connect FB directly to GND to select the fixed -4.1V output (Figure 2a). To select an alternate output voltage, connect FB to the midpoint of a resistor voltage divider from OUT to GND (Figure 2b).  $V_{IN}$  must be 1.0V above the absolute value of  $V_{OUT}$  to allow proper regulation. The output voltage is calculated from the formula below. Choose R2 to be between 100k $\Omega$  to 400k $\Omega$ .

 $V_{OUT} = (-1.28) \left( 1 + \frac{R^2}{R^1} \right)$ 





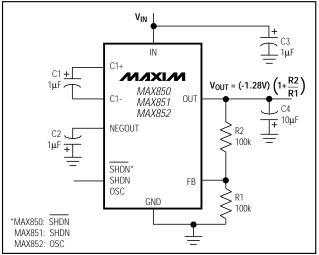
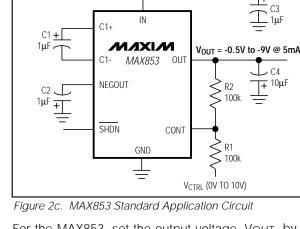


Figure 2b. MAX850/MAX851/MAX852 Adjustable Configuration



 $V_{\text{IN}}$ 

For the MAX853, set the output voltage, V<sub>OUT</sub>, by connecting a resistor voltage divider between OUT and a positive control voltage, V<sub>CTRL</sub> (Figure 2c).

$$V_{OUT} = -V_{CTRL} \left( \frac{R2}{R1} \right)$$

#### Shutdown

The MAX850–MAX853 feature a shutdown mode that reduces the supply current to 1 $\mu$ A max over temperature (5 $\mu$ A max for the MAX851). The MAX850 and MAX853 have an active-low TTL logic level SHDN input, whereas the MAX851 has an active-high SHDN input. To shut down the MAX852, set the OSC input to a logic-low level. The device is powered up by the resumption of the clock signal.

#### Capacitors

Use capacitors with low effective series resistance (ESR) to maintain a low dropout voltage ( $V_{IN} - |V_{OUT}|$ ). The overall dropout voltage is a function of the charge pump's output resistance and the voltage drop across the linear regulator (N-channel pass transistor). At the 100kHz switching frequency, the charge-pump output resistance is a function of C1 and C2's ESR. Therefore, minimizing the ESR of the charg-pump capacitors minimizes the dropout voltage.

1 $\mu$ F, 0.8 $\Omega$  ESR capacitors are recommended for C1, C2, and C3. C4 should be 10 $\mu$ F, 0.2 $\Omega$  ESR. All capacitors should be either surface-mount chip tantalum or chip ceramic types. External capacitor values may be adjusted to optimize size and cost.

#### Switching-Frequency Control

Use the MAX852 to minimize system interference caused by conflicting clock frequencies. An external oscillator can set the charge-pump frequency and reduce clock frequency sensitivity and interference. The clock must be



a square wave between 40% and 60% duty cycle. The maximum clock frequency is 250kHz and the minimum frequency is 50kHz.

#### Layout and Grounding

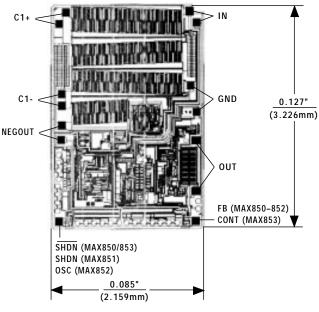
Good layout is important, primarily for good noise performance.

- 1) Mount all components as close together as possible.
- 2) Keep traces short to minimize parasitic inductance and capacitance. This includes connections to FB.
- 3) Use a ground plane.

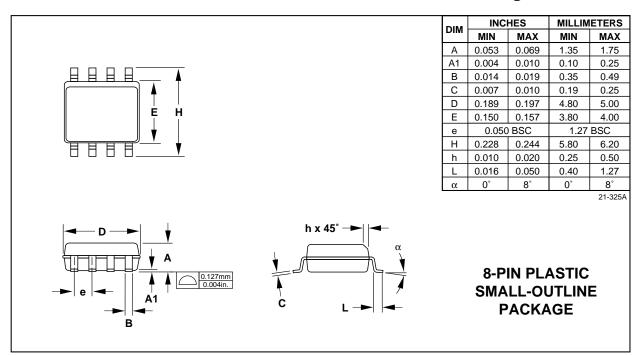
#### Noise and Ripple Measurement

Accurately measuring the output noise and ripple is a challenge. Brief differences in ground potential between the MAX850–MAX853 circuit and the oscilloscope (which result from the charge pump's switching action) cause ground currents in the probe's wires, inducing sharp voltage spikes. For best results, measure directly across the output capacitor (C4). Do not use the ground lead of the oscilloscope probe; instead, remove the probe's tip cover and touch the ground ring on the probe directly to C4's ground terminal. You can also use a Tektronix chassis mount test jack (part no. 131-0258) to connect your scope probe directly. This direct connection gives the most accurate noise and ripple measurement.

# \_Chip Topography



TRANSISTOR COUNT: 164 SUBSTRATE CONNECTED TO IN



## Package Information