

400MHz, Ultra-Low-Distortion Op Amps

General Description

The MAX4108/MAX4109 op amps combine ultra-high-speed performance with low-distortion operation. The MAX4108 is compensated for unity-gain stability, while the MAX4109 is compensated for a closed-loop gain (A_{vCL}) of 2V/V or greater.

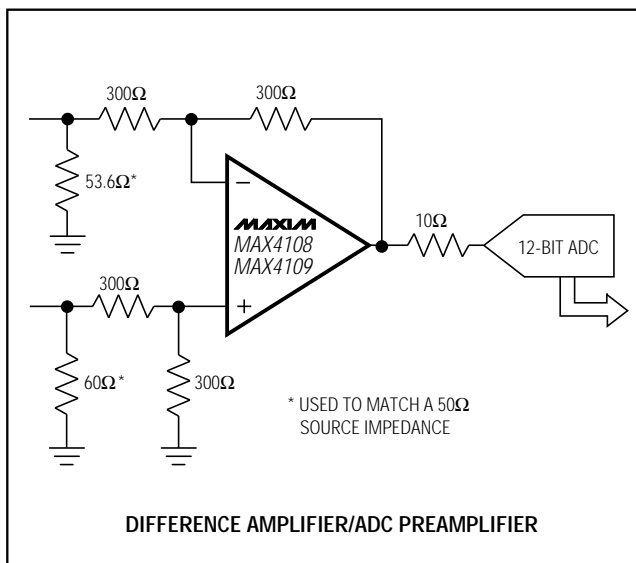
The MAX4108 delivers a 400MHz unity-gain bandwidth with a 1200V/ μ s slew rate, while the MAX4109 provides a 225MHz gain bandwidth with a 1200V/ μ s slew rate. The low-distortion design provides an unprecedented spurious-free dynamic range of -81dBc and -80dBc, respectively, at 20MHz ($V_O = 2V_{p-p}$, $R_L = 100\Omega$), making these amplifiers ideal for high-performance RF signal processing.

These high-speed op amps have a wide output voltage swing and a high current-drive capability of 90mA.

Applications

RGB and Composite Video
High-Performance Receivers
Pulse/RF Amplifier
ADC/DAC Preamp
Active Filters

Typical Application Circuit



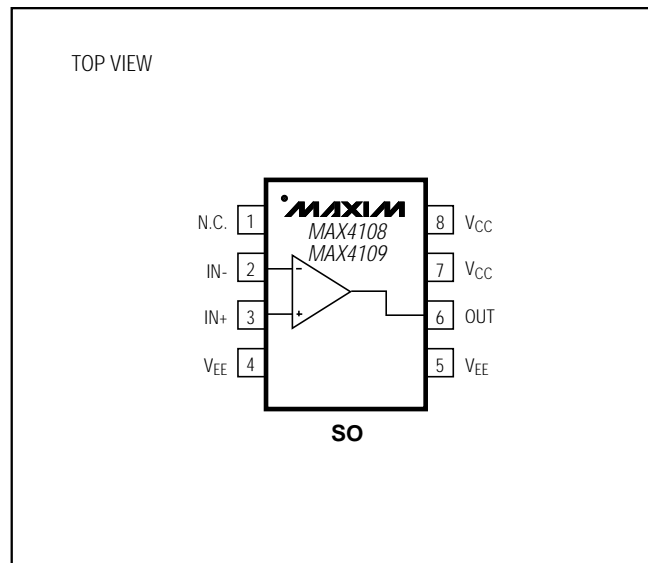
Features

- ◆ 400MHz Unity-Gain Bandwidth (MAX4108)
225MHz -3dB Bandwidth (MAX4109)
- ◆ 1200V/ μ s Slew Rate
- ◆ -81dBc SFDR at $f_C = 20\text{MHz}$ (MAX4108)
-80dBc SFDR at $f_C = 20\text{MHz}$ (MAX4109)
- ◆ 300MHz Full-Power Bandwidth (MAX4108, $V_O = 2V_{p-p}$)
200MHz Full-Power Bandwidth (MAX4109, $V_O = 2V_{p-p}$)
- ◆ Voltage Feedback Architecture
- ◆ High Output Drive: 90mA
- ◆ Output Short-Circuit Protected

Ordering Information

PART	TEMP. RANGE	PIN-PACKAGE
MAX4108ESA	-40°C to +85°C	8 SO
MAX4109ESA	-40°C to +85°C	8 SO

Pin Configuration



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MAX4108/MAX4109

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V_{CC} to V_{EE})12V
 Voltage on Any Pin to Ground or Any Other Pin V_{CC} to V_{EE}
 Short-Circuit Duration (V_{OUT} to GND)Continuous
 Continuous Power Dissipation ($T_A = +70^\circ\text{C}$)
 SO (derate 5.88mW/ $^\circ\text{C}$ above $+70^\circ\text{C}$)471mW

Operating Temperature Range
 MAX4108ESA/MAX4109ESA -40°C to $+85^\circ\text{C}$
 Storage Temperature Range -65°C to $+160^\circ\text{C}$
 Junction Temperature $+150^\circ\text{C}$
 Lead Temperature (soldering, 10sec) $+300^\circ\text{C}$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

($V_{CC} = 5\text{V}$, $V_{EE} = -5\text{V}$, $T_A = T_{MIN}$ to T_{MAX} , typical values are at $T_A = +25^\circ\text{C}$, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
DC SPECIFICATIONS							
Input Offset Voltage	V_{OS}	$V_{OUT} = 0\text{V}$			1	8	mV
Input Offset Voltage Drift	TCV_{OS}	$V_{OUT} = 0\text{V}$			13		$\mu\text{V}/^\circ\text{C}$
Input Bias Current	I_B	$V_{OUT} = 0\text{V}$, $V_{IN} = -V_{OS}$			12	25	μA
Input Offset Current	I_{OS}	$V_{OUT} = 0\text{V}$, $V_{IN} = -V_{OS}$			0.05	1.0	μA
Common-Mode Input Resistance	R_{INCM}	Either input			1.5		$\text{M}\Omega$
Common-Mode Input Capacitance	C_{INCM}	Either input			1		pF
Input Voltage Noise	e_n	$f_O = 10\text{kHz}$	MAX4108		8		$\text{nV}/\sqrt{\text{Hz}}$
			MAX4109		6		
Integrated Voltage Noise	E_{nRMS}	$f_B = 1\text{MHz to } 100\text{MHz}$	MAX4108		100		μVRMS
			MAX4109		75		
Input Current Noise	I_n	$f_O = 10\text{kHz}$	MAX4108		2		$\text{pA}/\sqrt{\text{Hz}}$
			MAX4109		2		
Integrated Current Noise	I_{nRMS}	$f_B = 1\text{MHz to } 100\text{MHz}$	MAX4108		25		nARMS
			MAX4109		25		
Common-Mode Input Voltage	V_{CM}			-2.5		2.5	V
Common-Mode Rejection	CMR	$V_{CM} = \pm 2.5\text{V}$		70	100		dB
Power-Supply Rejection	PSR	$V_S = \pm 4.5\text{V to } \pm 5.5\text{V}$		70	90		dB
Open-Loop Voltage Gain	A_{OL}	$V_{OUT} = \pm 2.0\text{V}$, $V_{CM} = 0\text{V}$, $R_L = 100\Omega$		70	100		dB
Quiescent Supply Current	I_{SY}	$V_{IN} = 0\text{V}$			20	25	mA
Output Voltage Swing	V_{OUT}	$R_L = \infty$		2.5 to -3.1	2.9 to -3.8		V
		$R_L = 100\Omega$		2.5 to -3.1	2.7 to -3.7		
Output Current Drive	I_{OUT}	$R_L = 33\Omega$, $T_A = 0^\circ\text{C to } +85^\circ\text{C}$		65	90		mA
Short-Circuit Output Current	I_{SC}	Short to ground			100		mA

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MAX4108/MAX4109

ELECTRICAL CHARACTERISTICS (continued)

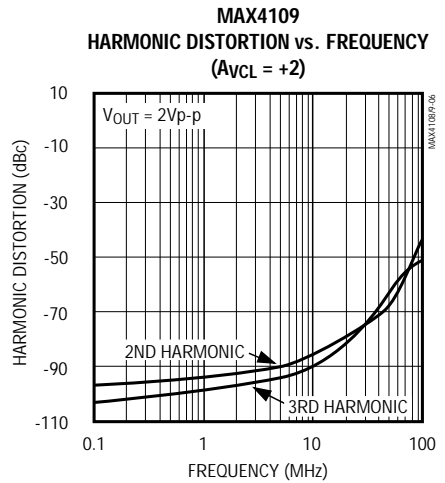
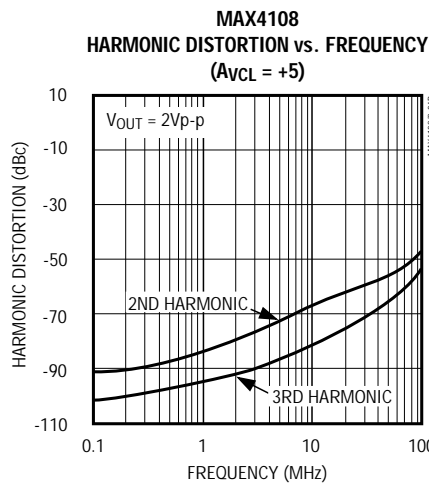
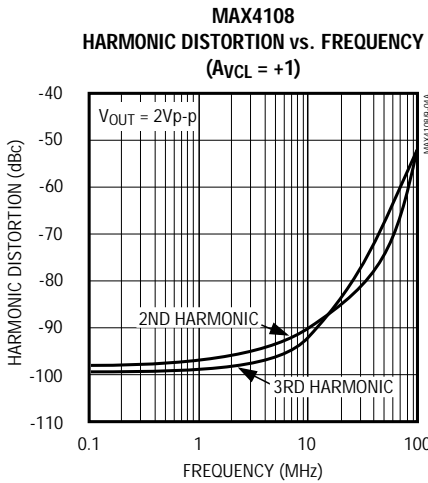
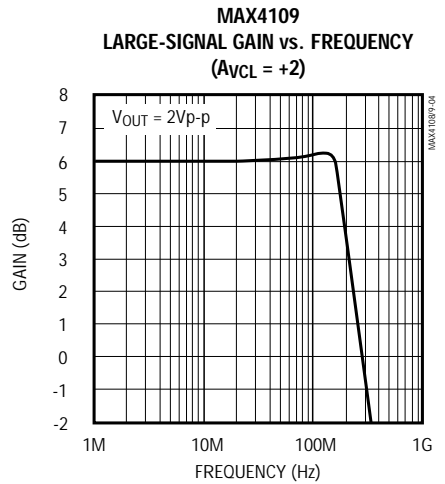
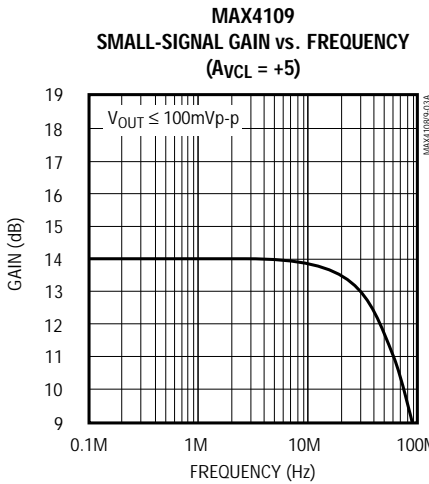
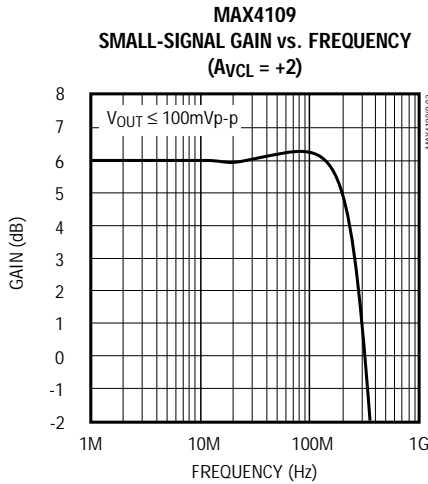
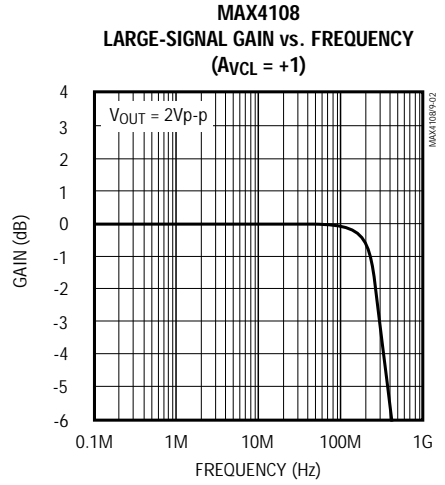
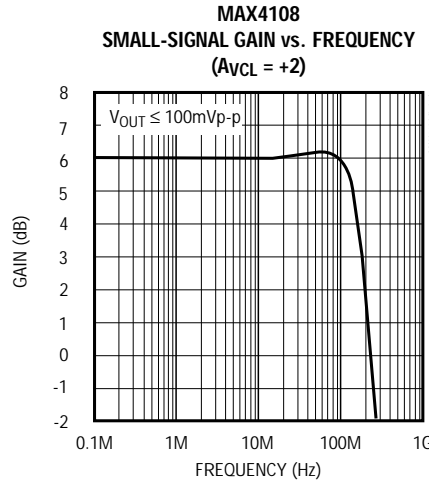
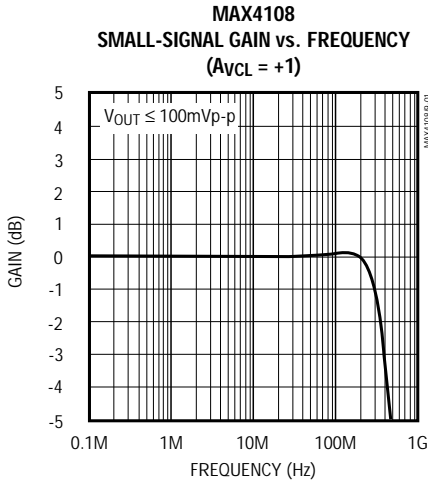
(V_{CC} = 5V, V_{EE} = -5V, T_A = T_{MIN} to T_{MAX}, typical values are at T_A = +25°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
AC SPECIFICATIONS							
-3dB Bandwidth	BW _{-3dB}	V _{OUT} ≤ 0.1V _{RMS}	MAX4108		400		MHz
			MAX4109		225		
Full-Power Bandwidth	FPBW	V _{OUT} = 2V _{p-p}	MAX4108		300		MHz
			MAX4109		200		
0.1dB Bandwidth	BW _{0.1dB}	MAX4108, A _{VCL} = +1			100		MHz
		MAX4109, A _{VCL} = +2			25		
Slew Rate	SR	-2V ≤ V _{OUT} ≤ 2V	MAX4108		1200		V/μs
			MAX4109		1200		
Settling Time	t _s	to 0.1%, -1V ≤ V _{OUT} ≤ 1V, R _L = 100Ω	MAX4108		8		ns
			MAX4109		8		
		to 0.01%, -1V ≤ V _{OUT} ≤ 1V, R _L = 100Ω	MAX4108		12		
			MAX4109		12		
Rise/Fall Times	t _R , t _F	-2V ≤ V _{OUT} ≤ 2V, 10% to 90%, R _L = 100Ω	MAX4108		3		ns
			MAX4109		3		
		-50mV ≤ V _{OUT} ≤ 50mV, 10% to 90%, R _L = 100Ω	MAX4108		2		
			MAX4109		2		
Differential Gain	DG	f = 3.58MHz	MAX4108, A _{VCL} = +1		0.002		%
			MAX4109, A _{VCL} = +2		0.008		
Differential Phase	DP	f = 3.58MHz	MAX4108, A _{VCL} = +1		0.002		degrees
			MAX4109, A _{VCL} = +2		0.002		
Input Capacitance	C _{IN}				2		pF
Output Resistance	R _{OUT}	f = 10MHz			1		Ω
Spurious-Free Dynamic Range	SFDR	MAX4108, A _{VCL} = +1, V _{OUT} = 2V _{p-p}	f _C = 5MHz		-93		dBc
			f _C = 20MHz		-81		
		MAX4109, A _{VCL} = +2, V _{OUT} = 2V _{p-p}	f _C = 5MHz		-90		
			f _C = 20MHz		-80		
Third-Order Intercept	IP3	f _C = 10MHz	MAX4108, A _{VCL} = +1		39		dBm
			MAX4109, A _{VCL} = +2		36		

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Typical Operating Characteristics

($V_{CC} = +5V$, $V_{EE} = -5V$, $R_L = 100\Omega$, $T_A = +25^\circ C$, unless otherwise noted.)

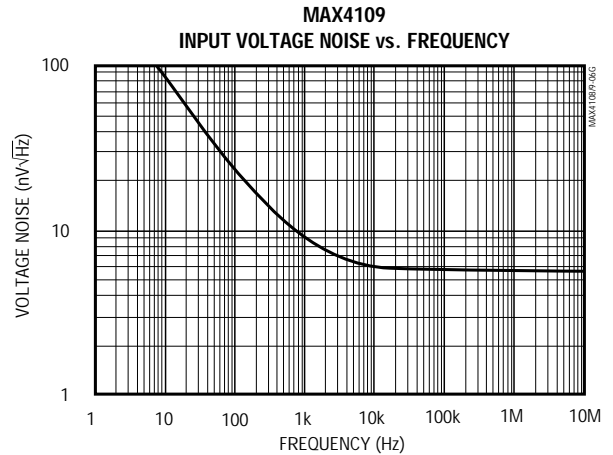
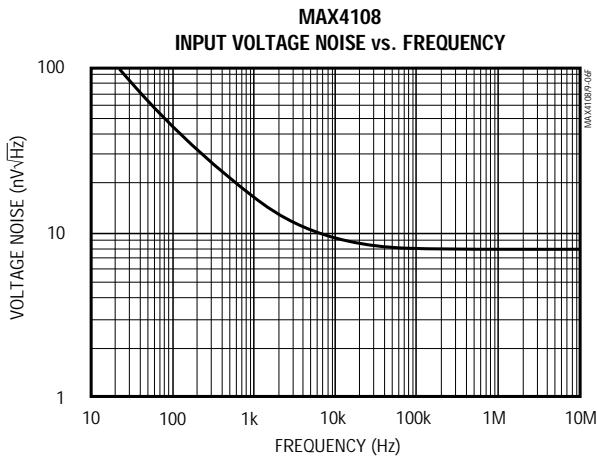
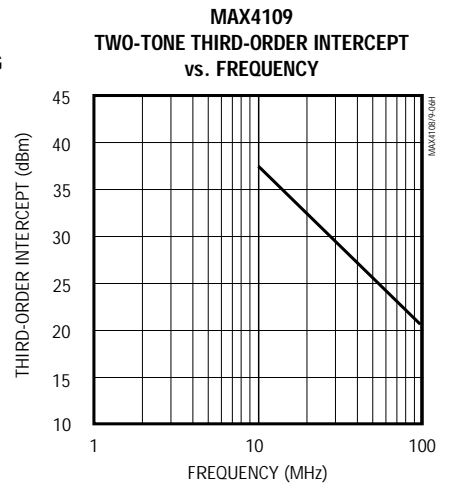
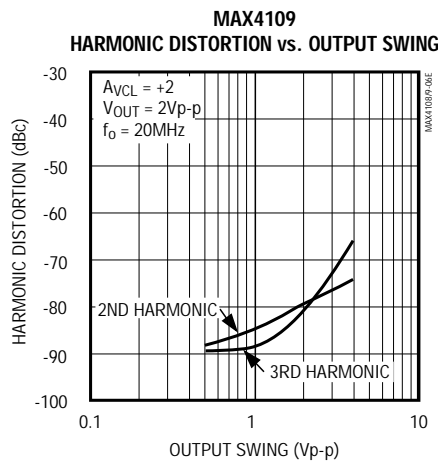
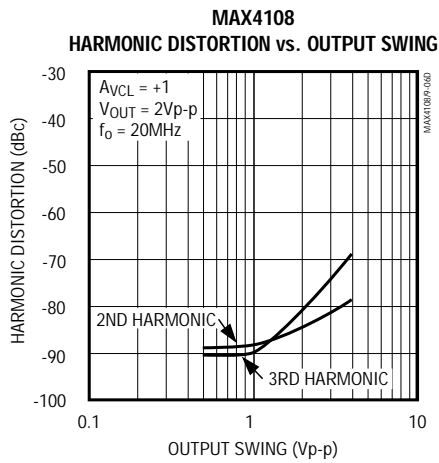
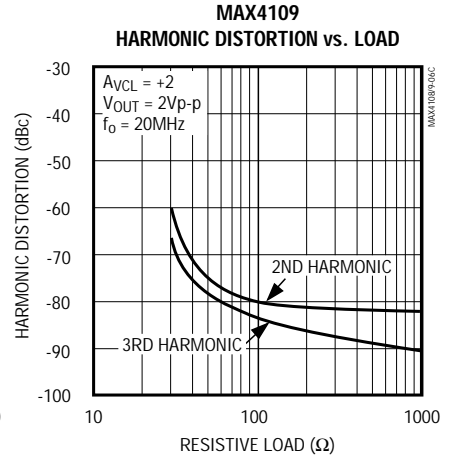
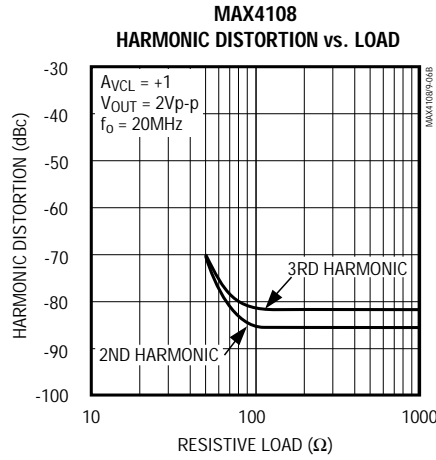
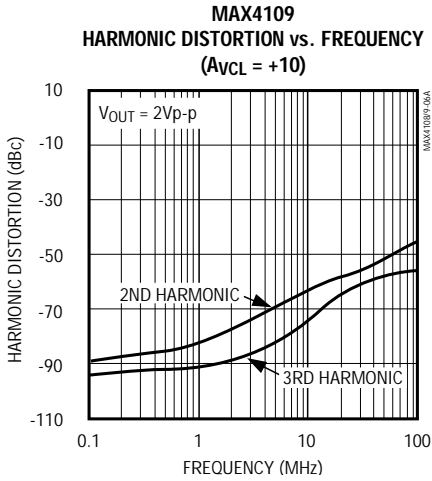


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Typical Operating Characteristics (continued)

($V_{CC} = +5V$, $V_{EE} = -5V$, $R_L = 100\Omega$, $T_A = +25^\circ C$, unless otherwise noted.)

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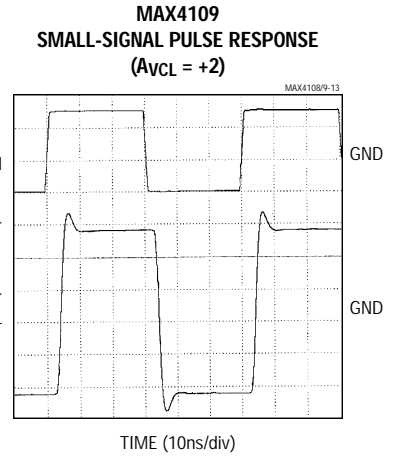
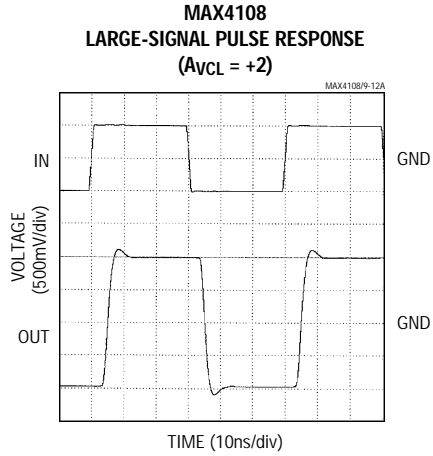
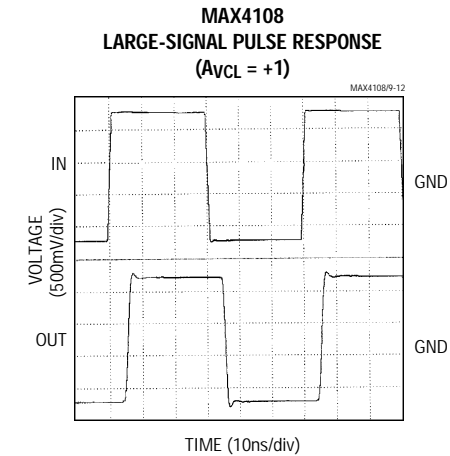
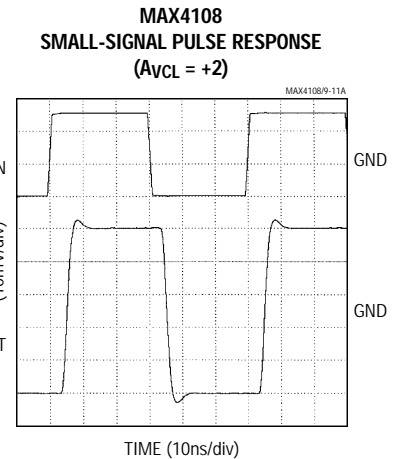
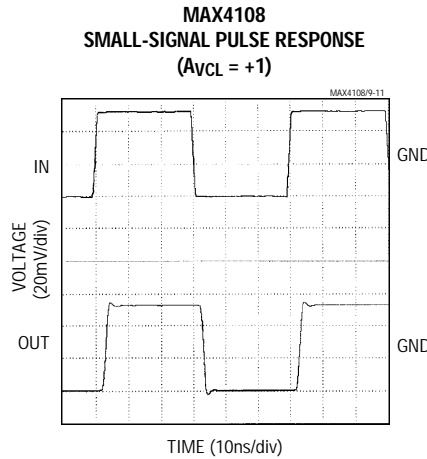
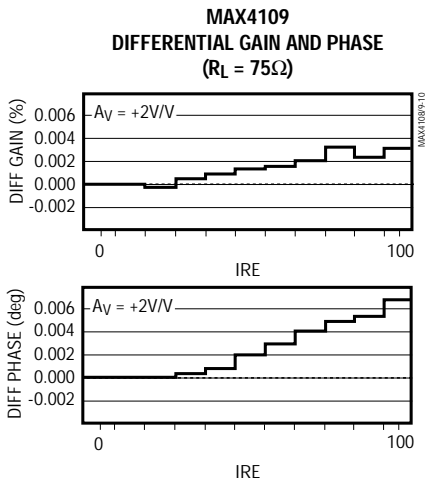
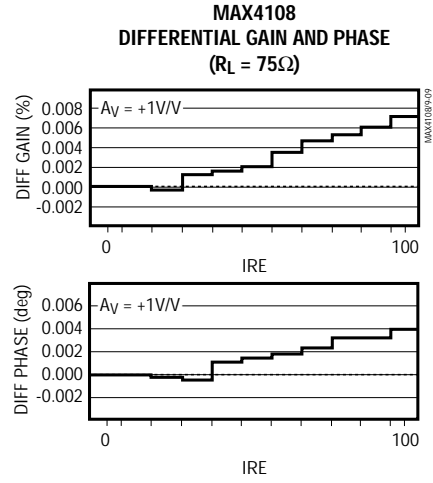
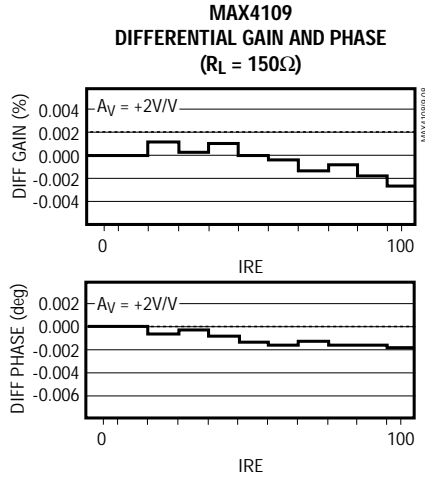
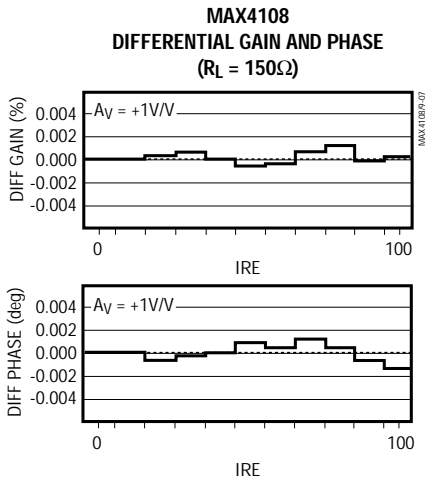


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MAX4108/MAX4109

Typical Operating Characteristics (continued)

($V_{CC} = +5V$, $V_{EE} = -5V$, $R_L = 100\Omega$, $T_A = +25^\circ C$, unless otherwise noted.)

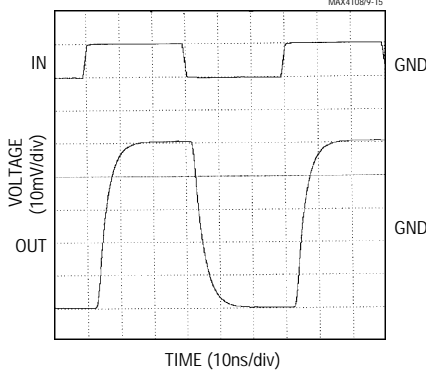


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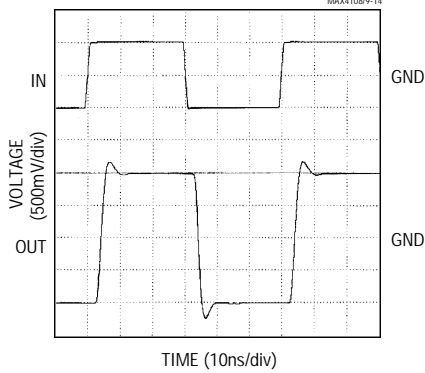
Typical Operating Characteristics (continued)

MAX4108/MAX4109

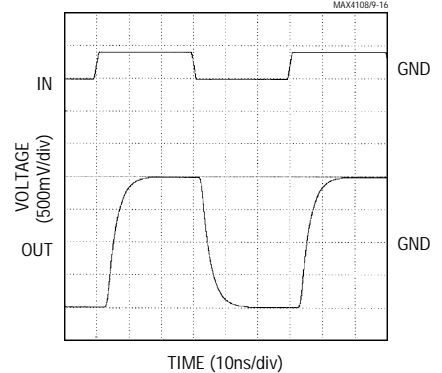
**MAX4109
SMALL-SIGNAL PULSE RESPONSE
($A_{VCL} = +5$)**



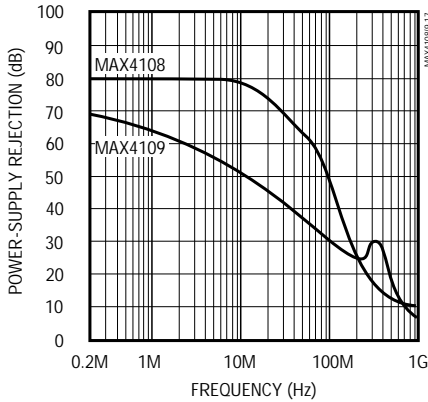
**MAX4109
LARGE-SIGNAL PULSE RESPONSE
($A_{VCL} = +2$)**



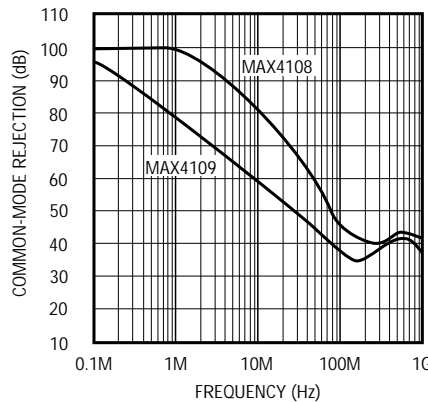
**MAX4109
LARGE-SIGNAL PULSE RESPONSE
($A_{VCL} = +5$)**



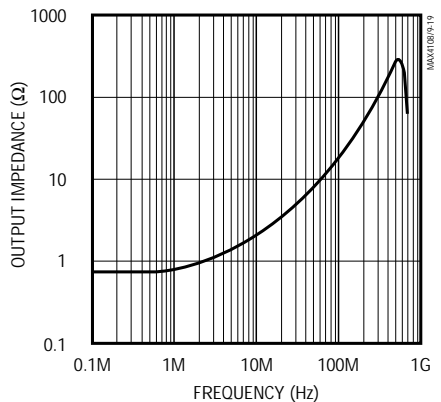
**POWER-SUPPLY REJECTION
vs. FREQUENCY**



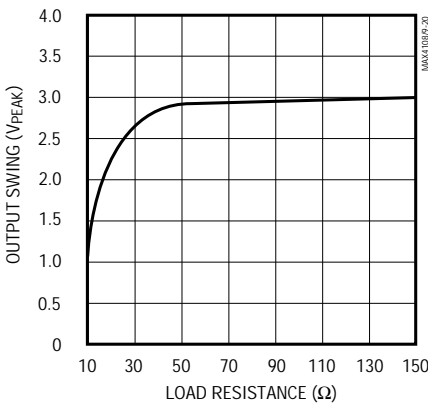
**COMMON-MODE REJECTION
vs. FREQUENCY**



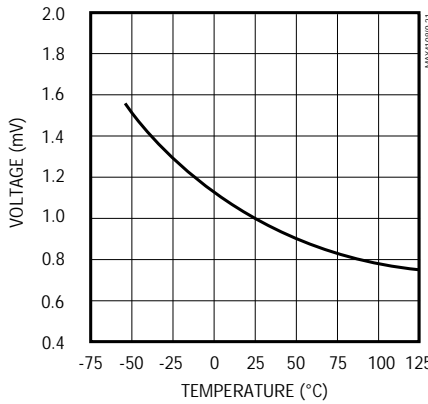
**CLOSED-LOOP OUTPUT IMPEDANCE
vs. FREQUENCY**



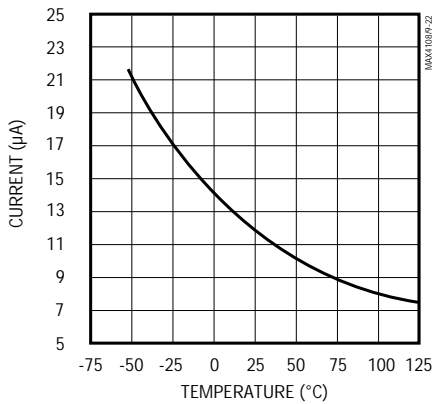
**MAX4108
OUTPUT SWING vs. LOAD RESISTANCE**



**INPUT OFFSET VOLTAGE
vs. TEMPERATURE**



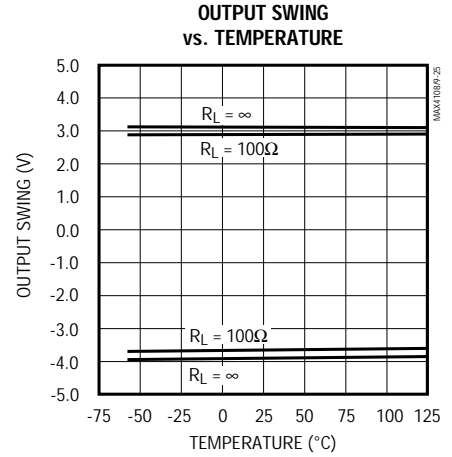
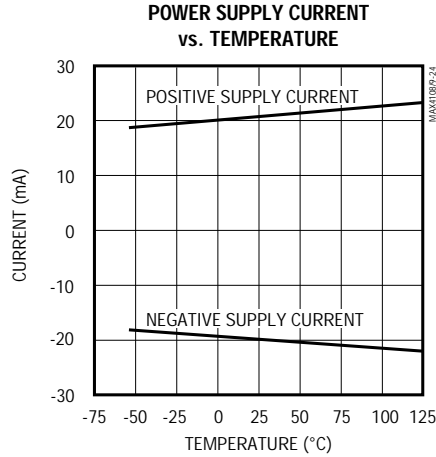
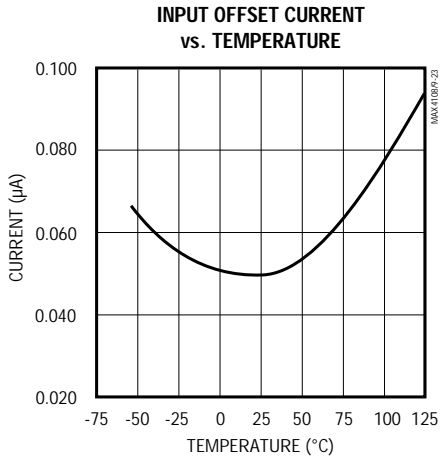
**INPUT BIAS CURRENT
vs. TEMPERATURE**



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Typical Operating Characteristics (continued)

($V_{CC} = +5V$, $V_{EE} = -5V$, $R_L = 100\Omega$, $T_A = +25^\circ C$, unless otherwise noted.)



Pin Description

PIN	NAME	FUNCTION
1	N.C.	No Connection. Not internally connected.
2	IN-	Inverting Input
3	IN+	Noninverting Input
4, 5	V_{EE}	Negative Power Supply, connect to -5V
6	OUT	Amplifier Output
7, 8	V_{CC}	Positive Power Supply, connect to +5V

Detailed Description

Choosing Resistor Values

Unity-Gain Configuration

The MAX4108 is internally compensated for unity gain. When configured for unity gain, the device requires a small resistor (typically 24Ω) in series with the feedback path. This resistor provides better AC response by reducing the Q of the tank circuit formed by parasitic feedback inductance and capacitance.

Inverting and Noninverting Configurations

The values of the gain-setting feedback and input resistors are important design considerations. Large resistor values will increase voltage noise, and will interact with the amplifier's input and PC board capacitance to generate undesirable poles and zeros, which can decrease bandwidth or cause oscillations. For example, a noninverting gain of +2, using $1k\Omega$ resistors combined with $2pF$ of input capacitance and $0.5pF$ of board capacitance, will cause a feedback pole at $128MHz$. If this pole is within the anticipated amplifier bandwidth, it will jeopardize stability. Reducing these $1k\Omega$ resistors to 100Ω will extend the pole frequency to $1.28GHz$, but could limit output swing by adding 200Ω in parallel with the amplifier's load. Clearly, the selection of resistor values must be tailored to the specific application.

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The MAX4108/MAX4109 are ultra-low-distortion, high-bandwidth op amps. The output distortion will be degraded as the total load resistance seen by the amplifier decreases. To minimize distortion products, keep the input and gain-setting resistors relatively large. A 500Ω feedback resistor combined with an appropriate input resistor to set the gain will provide excellent AC performance without significantly increasing distortion.

Driving Capacitive Loads

The MAX4108/MAX4109 are optimized for AC performance. They are not designed to drive highly reactive loads. Reactive loads will decrease phase margin and may produce excessive ringing and oscillation. Figure 1a shows a circuit that eliminates this problem, and Figure 1b is a graph of the optimal isolation resistor

(R_s) vs. capacitive load. Figures 2a and 2b show how a capacitive load causes excessive peaking of the amplifier's bandwidth if the capacitive load is not isolated (R_s) from the amplifier. A small isolation resistor (usually 10Ω to 20Ω) placed before the reactive load prevents ringing and oscillation. At higher capacitive loads, AC performance will be controlled by the interaction of the load capacitance and isolation resistor. Figures 3a and 3b show the effect of an isolation resistor on the MAX4108/MAX4109 closed-loop response.

Coaxial cable and other transmission lines are easily driven when terminated at both ends with their characteristic impedance. When driving back-terminated transmission lines, the capacitance of the transmission line is essentially eliminated.

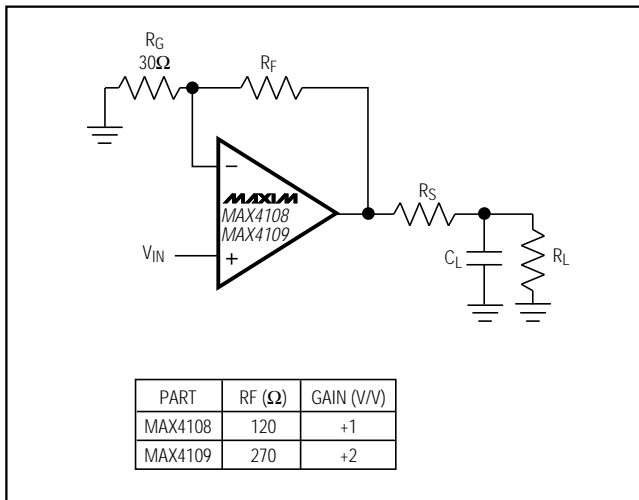


Figure 1a. Using an Isolation Resistor for High Capacitive Loads

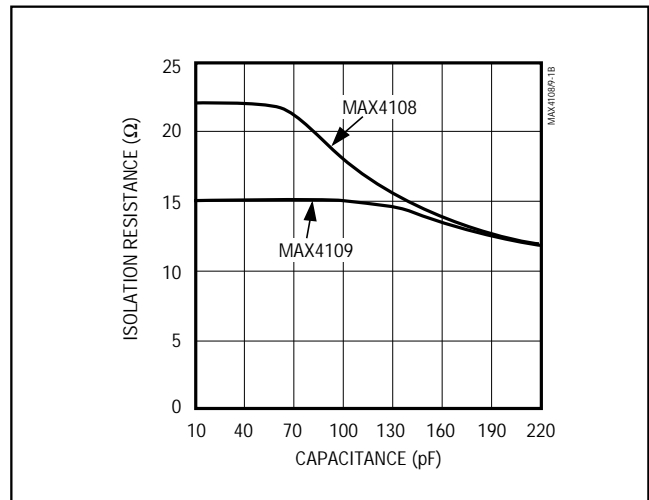


Figure 1b. Optimal Isolation Resistor (R_s) vs. Capacitive Load

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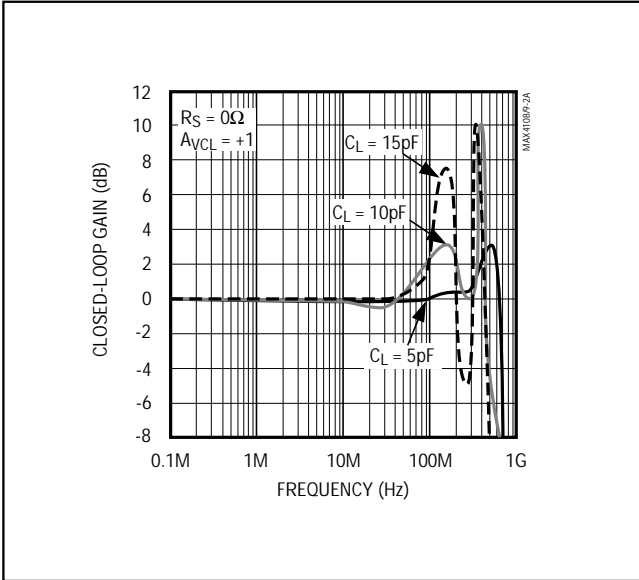


Figure 2a. MAX4108 Response vs. Capacitive Load—No Resistive (R_S) Isolation (circuit shown in Figure 1)

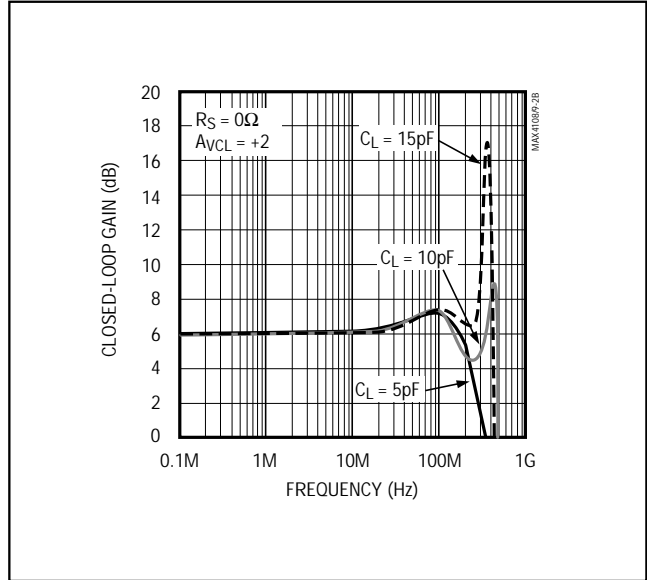


Figure 2b. MAX4109 Response vs. Capacitive Load—No Isolation (R_S) Resistor (circuit shown in Figure 1)

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MAX4108/MAX4109

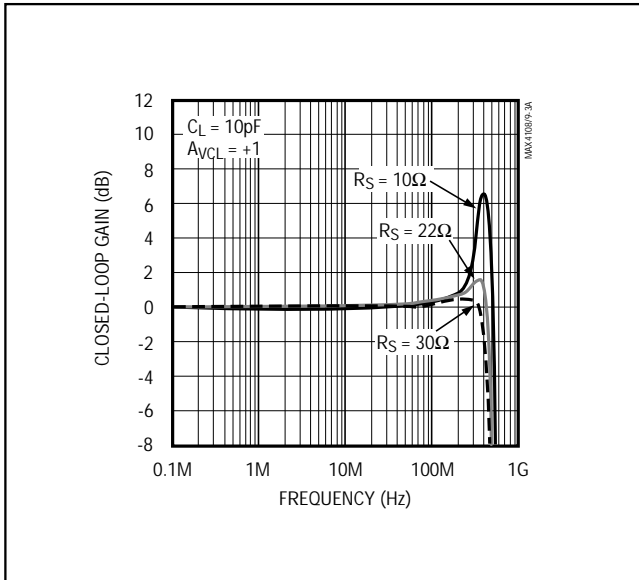


Figure 3a. MAX4108 Response vs. Capacitive Load with Resistive (R_S) Isolation (circuit shown in Figure 1)

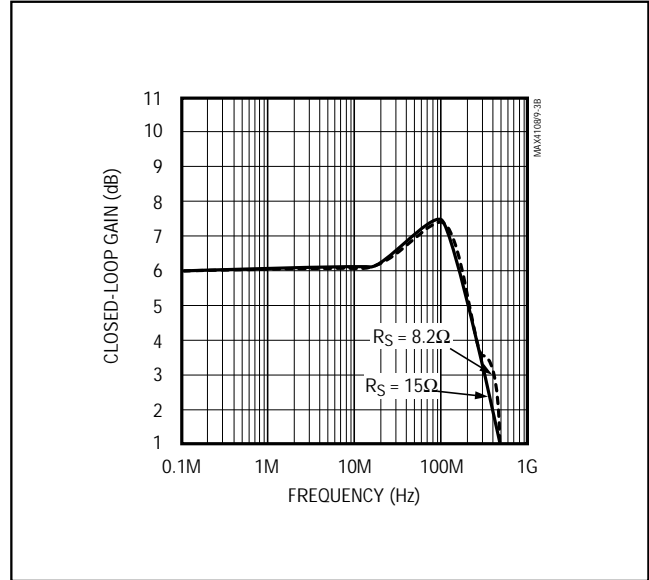


Figure 3b. MAX4109 Response vs. Capacitive Load with Resistive (R_S) Isolation (circuit shown in Figure 1)

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Chip Information

TRANSISTOR COUNT: 57
 SUBSTRATE CONNECTED TO VEE

Package Information

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.053	0.069	1.35	1.75
A1	0.004	0.010	0.10	0.25
B	0.014	0.019	0.35	0.49
C	0.007	0.010	0.19	0.25
E	0.150	0.157	3.80	4.00
e	0.050		1.27	
H	0.228	0.244	5.80	6.20
L	0.016	0.050	0.40	1.27

DIM	PINS	INCHES		MILLIMETERS	
		MIN	MAX	MIN	MAX
D	8	0.189	0.197	4.80	5.00
D	14	0.337	0.344	8.55	8.75
D	16	0.386	0.394	9.80	10.00

21-0041A

**Narrow SO
 SMALL-OUTLINE
 PACKAGE
 (0.150 in.)**

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

12 Maxim Integrated Products, 120 San Gabriel Drive, Sunnyvale, CA 94086 (408) 737-7600