## 200-MHz Field Programmable Zero Delay Buffer

## Features

- Pre-programmed configuration

■ Fully field-programmable $\square$ Input and output dividers a Inverting/non inverting outputs
a Phase-locked loop (PLL) or fanout buffer configuration
■ 10 MHz to 200 MHz operating range
■ Split $2.5-\mathrm{V}$ or $3.3-\mathrm{V}$ outputs
■ Two low-voltage complementary metal oxide semiconductor (LVCMOS) reference inputs

■ Twelve low-skew outputs
a Output-output skew < 200 ps
a Device-device skew < 500 ps
■ Input-output skew < 250 ps
■ Cycle-cycle jitter < 100 ps (typical)

- Three-stateable outputs

■ Less than $50 \mu \mathrm{~A}$ shutdown current

- Spread Aware ${ }^{\text {TM }}$

■ 28-pin shrunk small outline package (SSOP)
-3.3-V operation

## Functional Description

The CY23FP12-002 is a pre-programmed version of the CY23FP12. It features a high-performance fully field-programmable $200-\mathrm{MHz}$ zero delay buffer designed for high-speed clock distribution. The integrated PLL is designed for low jitter and optimized for noise rejection. These parameters are critical for reference clock distribution in systems using high-performance ASICs and microprocessors.
The CY23FP12-002 is fully programmable through volume or prototype programmers, enabling the user to define an application-specific zero delay buffer with customized input and output dividers, feedback topology (internal/external), output inversions, and output drive strengths. For additional flexibility, the user can mix and match multiple functions, listed in Table 2 on page 5, and assign a particular function set to any one of the four possible S1-S2 control bit combinations. This feature enables the implementation of four distinct personalities, selectable with S1-S2 bits, on a single programmed silicon. The CY23FP12-002 also features a proprietary auto power down circuit that shuts down the device in case of a REF failure, resulting in less than $50 \mu \mathrm{~A}$ of current draw.
The CY23FP12-002 provides 12 outputs grouped in two banks with separate power supply pins which can be connected independently to either a 2.5 V or a 3.3 V rail.
Selectable reference input is a fault tolerance feature which allows for glitch-free switch over to secondary clock source when REFSEL is asserted/deasserted.

## Logic Block Diagram



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## Pin Configuration

Figure 1. 28-Pin SSOP

| REF2 $\square 1$ | 28 | REFSEL |
| :---: | :---: | :---: |
| REF1 $\square 2$ | 27 | FBK |
| CLKBO $\square$ | 26 | CLKAO |
| CLKB1 $\square$ | 25 | CLKA1 |
| $\mathrm{V}_{\text {SSB }} \square 5$ | 24 | $\mathrm{V}_{\text {SSA }}$ |
| CLKB2 $\square$ | 23 | CLKA2 |
| CLKB3 $\square$ | 22 | CLKA3 |
| $V_{\text {DDB }} \square$ | 21 | $V_{\text {DDA }}$ |
| $\mathrm{V}_{\text {SSB }} \square$ | 20 | $V_{\text {SSA }}$ |
| CLKB4 $\square^{10}$ | 19 | CLKA4 |
| CLKB5 $\square 11$ | 18 | CLKA5 |
| $\mathrm{V}_{\text {DDB }} \square{ }^{12}$ | 17 | $V_{\text {DDA }}$ |
| $\mathrm{V}_{\text {DDC }} \square^{13}$ | 16 | $\mathrm{V}_{\text {SSC }}$ |
| S2 $\square 14$ | 15 | S1 |

## Pin Description

| Pin | Name | 1/0 | Type | Description |
| :---: | :---: | :---: | :---: | :---: |
| 1 | REF2 | I | LVTTL/LVCMOS | Input reference frequency, 5 V -tolerant input. |
| 2 | REF1 | 1 | LVTTL/LVCMOS | Input reference frequency, 5 V -tolerant input. |
| 3 | CLKB0 | 0 | LVTTL | Clock output, Bank B. |
| 4 | CLKB1 | 0 | LVTTL | Clock output, Bank B. |
| 5 | $\mathrm{V}_{\text {SSB }}$ | PWR | POWER | Ground for Bank B. |
| 6 | CLKB2 | 0 | LVTTL | Clock output, Bank B. |
| 7 | CLKB3 | O | LVTTL | Clock output, Bank B. |
| 8 | $V_{\text {DDB }}$ | PWR | POWER | 2.5-V or 3.3-V supply, Bank B. |
| 9 | $\mathrm{V}_{\text {SSB }}$ | PWR | POWER | Ground for Bank B. |
| 10 | CLKB4 | O | LVTTL | Clock output, Bank B. |
| 11 | CLKB5 | O | LVTTL | Clock output, Bank B. |
| 12 | $V_{\text {DDB }}$ | PWR | POWER | 2.5-V or 3.3-V supply, Bank B. |
| 13 | $\mathrm{V}_{\text {DDC }}$ | PWR | POWER | 3.3 V core supply. |
| 14 | S2 | I | LVTTL | Select input. |
| 15 | S1 | I | LVTTL | Select input. |
| 16 | $\mathrm{V}_{\text {SSC }}$ | PWR | POWER | Ground for core. |
| 17 | V ${ }_{\text {DDA }}$ | PWR | POWER | 2.5-V or 3.3-V supply, Bank A. |
| 18 | CLKA5 | O | LVTTL | Clock output, Bank A. |
| 19 | CLKA4 | 0 | LVTTL | Clock output, Bank A. |
| 20 | $\mathrm{V}_{\text {SSA }}$ | PWR | POWER | Ground for Bank A. |
| 21 | $\mathrm{V}_{\text {DDA }}$ | PWR | POWER | $2.5-\mathrm{V}$ or 3.3-V supply Bank A. |
| 22 | CLKA3 | O | LVTTL | Clock output, Bank A. |
| 23 | CLKA2 | O | LVTTL | Clock output, Bank A. |
| 24 | $\mathrm{V}_{\text {SSA }}$ | PWR | POWER | Ground for Bank A. |
| 25 | CLKA1 | O | LVTTL | Clock output, Bank A. |
| 26 | CLKA0 | 0 | LVTTL | CLock output, Bank A. |
| 27 | FBK | 1 | LVTTL | PLL feedback input. |
| 28 | REFSEL | 1 | LVTTL | Reference select input. When REFSEL = 0, REF1 is selected. When REFSEL = 1, REF2 is selected. |

Figure 2. Basic PLL Block Diagram


Following is a list of independent functions that can be programmed with a volume or prototype programmer on the "pre-programmed" silicon.

Table 1. Programmable Functions

| Configuration | Description | Default |
| :--- | :--- | :--- |
| DC Drive Bank A | Programs the drive strength of Bank A outputs. The user can select one out of <br> two possible drive strength settings that produce output DC currents in the range <br> of $\pm 16 \mathrm{~mA}$ to $\pm 20 \mathrm{~mA}$. | $\pm 20 \mathrm{~mA}$ |
| DC Drive Bank B | Programs the drive strength of Bank B outputs. The user can select one out of <br> two possible drive strength settings that produce output DC currents in the range <br> of $\pm 16 \mathrm{~mA}$ to $\pm 20 \mathrm{~mA}$. | $\pm 20 \mathrm{~mA}$ |
| Output Enable for Bank B clocks | Enables/disables CLKB[5:0] outputs. Each of the six outputs can be disabled <br> individually if not used, to minimize electromagnetic interference (EMI) and <br> switching noise. | Enable |
| Output Enable for Bank A clocks | Enables/disables CLKA[5:0] outputs. Each of the six outputs can be disabled <br> individually if not used, to minimize EMI and switching noise. | Enable |
| Inv CLKA0 | Generates an inverted clock on the CLKAO output. When this option is <br> programmed, CLKA0 and CLKA1 will become complimentary pairs. | Non invert |
| Inv CLKA2 | Generates an inverted clock on the CLKA2 output. When this option is <br> programmed, CLKA2 and CLKA3 will become complimentary pairs. | Non invert |
| Inv CLKA4 | Generates an inverted clock on the CLKA4 output. When this option is <br> programmed, CLKA4 and CLKA5 will become complimentary pairs. | Non invert |
| Inv CLKB0 | Generates an inverted clock on the CLKB0 output. When this option is <br> programmed, CLKB0 and CLKB1 will become complimentary pairs. | Non invert |

Table 1. Programmable Functions (continued)

| Configuration | Description | Default |
| :--- | :--- | :--- |
| Inv CLKB2 | Generates an inverted clock on the CLKB2 output. When this option is <br> programmed, CLKB2 and CLKB3 will become complimentary pairs. | Non-invert |
| Inv CLKB4 | Generates an inverted clock on the CLKB4 output. When this option is <br> programmed, CLKB4 and CLKB5 will become complimentary pairs. | Non-invert |
| Pull down Enable | Enables/disables internal pulldowns on all outputs | Enable |
| Fbk Pull down Enable | Enables/disables internal pulldowns on the feedback path (applicable to both <br> internal and external feedback topologies) | Enable |
| Fbk Sel | Selects between the internal and the external feedback topologies | Internal |

The following table lists independent functions, which can be assigned to each of the four S1 and S2 combinations. When a particular S1 and S2 combination is selected, the device assumes the configuration (which is essentially a set of functions given in Table 2) that has been preassigned to that particular combination.
Table 2. Programmable Functions for S1/S2 Combinations

| Function | Description | Default |
| :---: | :---: | :---: |
| Output Enable CLKB[5:4] | Enables/disables CLKB[5:4] output pair | Enable |
| Output Enable CLKB[3:2] | Enables/disables CLKB[3:2] output pair | Enable |
| Output Enable CLKB[1:0] | Enables/disables CLKB[1:0] output pair | Enable |
| Output Enable CLKA[5:4] | Enables/disables CLKA[5:4] output pair | Enable |
| Output Enable CLKA[3:2] | Enables/disables CLKA[3:2] output pair | Enable |
| Output Enable CLKA[1:0] | Enables/disables CLKA[1:0] output pair | Enable |
| Auto Power down Enable | Enables/disables the auto power down circuit, which monitors the reference clock rising edges and shuts down the device in case of a reference 'failure.' This failure is triggered by a drift in reference frequency below a set limit. This auto power down circuit is disabled internally when one or more of the outputs are configured to be driven directly from the reference clock. | Enable |
| PLL Power down | Shuts down the PLL when the device is configured as a non-PLL fanout buffer. | See Table 4 on page 6 |
| M[7:0] | Assigns an eight-bit value to reference divider - M . The divider can be any integer value from 1 to 256; however, the PLL input frequency cannot be lower than 10 MHz . | See Table 4 on page 6 |
| N[7:0] | Assigns an eight-bit value to feedback divider - N . The divider can be any integer value from 1 to 256; however, the PLL input frequency cannot be lower than 10 MHz . | See Table 4 on page 6 |
| X[6:0] | Assigns a seven-bit value to output divider -X. The divider can be any integer value from 5 to 130 . Divide by $1,2,3$, and 4 are preprogrammed on the device and can be activated by the appropriate output mux setting. | See Table 4 on page 6 |
| Divider Source | Selects between the PLL output and the reference clock as the source clock for the output dividers. | See Table 4 on page 6 |
| CLKA54 Source | Independently selects one out of the eight possible output dividers that will connect to the CLKA5 and CLKA4 pair. Please refer to Table 3 on page 6 for a list of divider values. | See Table 4 on page 6 |
| CLKA32 Source | Independently selects one out of the eight possible output dividers that will connect to the CLKA3 and CLKA2 pair. Please refer to Table 3 on page 6 for a list of divider values. | See Table 4 on page 6 |
| CLKA10 Source | Independently selects one out of the eight possible output dividers that will connect to the CLKA1 and CLKAO pair. Please refer to Table 3 on page 6 for a list of divider values. | See Table 4 on page 6 |
| CLKB54 Source | Independently selects one out of the eight possible output dividers that will connect to the CLKB5 and CLKB4 pair. Please refer to Table 3 on page 6 for a list of divider values. | See Table 4 on page 6 |
| CLKB32 Source | Independently selects one out of the eight possible output dividers that will connect to the CLKB3 and CLKB2 pair. Please refer to Table 3 on page 6 for a list of divider values. | See Table 4 on page 6 |
| CLKB10 Source | Independently selects one out of the eight possible output dividers that will connect to the CLKB1 and CLKB0 pair. Please refer to Table 3 on page 6 for a list of divider values. | See Table 4 on page 6 |

Table 3 is a list of output dividers that are independently selected to connect to each output pair.
In the default (pre-programmed) state of the device, S1 and S2 pins will function as indicated in Table 4. The CY23FP12-002 can be programmed to other configurations.
Table 3. Output Dividers

| CLKA/B Source | Output Connects To |
| :---: | :---: |
| $0[000]$ | REF |
| $1[001]$ | Divide by 1 |
| $2[010]$ | Divide by 2 |
| $3[011]$ | Divide by 3 |
| $4[100]$ | Divide by 4 |
| $5[101]$ | Divide by $X$ |
| $6[110]$ | Divide by 2X ${ }^{[1]}$ |
| $7[111]$ | TEST mode [LOCK signal] ${ }^{[2]}$ |

Table 4. Pre-Programmed Configuration

| Outputs | S2, S1 | DivSrc | Example Output |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | REF Input (MHz) | VCO (MHz) | Output (MHz) |
| ClkA0, A1 | 00 | 1 | 25 | 200 | 200 |
| ClkA2, A3 | 00 | 3 | 25 | 200 | 66.7 |
| ClkA4, A5 | 00 | X=6 | 25 | 200 | 33.3 |
| ClkB0, B1 | 00 | X=6 | 25 | 200 | 33.3 |
| ClkB2, B3 | 00 | 4 | 25 | 200 | 50 |
| ClkB4, B5 | 00 | Ref | 25 | 200 | 25 |
| ClkA0, A1 | 01 | 4 | 100 | 200 | 50 |
| ClkA2, A3 | 01 | 4 | 100 | 200 | 50 |
| ClkA4, A5 | 01 | 4 | 100 | 200 | 50 |
| ClkB0, B1 | 01 | 4 | 100 | 200 | 50 |
| ClkB2, B3 | 01 | X=8 | 100 | 200 | 25 |
| ClkB4, B5 | 01 | X=8 | 100 | 200 | 25 |
| ClkA0, A1 | 10 | X=8 | 33.3 | 266.6 | 33.3 |
| ClkA2, A3 | 10 | X=8 | 33.3 | 266.6 | 33.3 |
| ClkA4, A5 | 10 | X=8 | 33.3 | 266.6 | 33.3 |
| ClkB0, B1 | 10 | 4 | 33.3 | 266.6 | 66.6 |
| ClkB2, B3 | 10 | 4 | 33.3 | 266.6 | 66.6 |
| ClkB4, B5 | 10 | 4 | 33.3 | 266.6 | 66.6 |
| ClkA0, A1 | 11 | Ref | 100 | powerdown | 100 |
| ClkA2, A3 | 11 | Ref | 100 | powerdown | 100 |
| ClkA4, A5 | 11 | Ref | 100 | powerdown | 100 |
| ClkB0, B1 | 11 | 2 | 100 | powerdown | 50 |
| ClkB2, B3 | 11 | 2 | 100 | powerdown | 50 |
| ClkB4, B5 | 11 | 2 | 100 | powerdown | 50 |

## Notes

1. Outputs will be rising edge aligned only to those outputs using this same device setting.
2. When the source of an output pair is set to [111], the output pair becomes lock indicator signal. For example, if the source of an output pair (CLKA0, CLKA1) is set to [111], the CLKA0 and CLKA1, becomes lock indicator signals. In non-invert mode, CLKA0 and CLKA1 signals will be high when the PLL is in lock mode. If CLKA0 is in an invert mode, the CLKA0 will be low and the CLKA1 will be high when the PLL is in lock mode.

## Field Programming the CY23FP12-002

The CY23FP12-002 comes pre-programmed and ready for use, but it can also be reprogrammed to any other valid configuration. When programming, it must be programmed in a device programmer prior to being installed in a circuit. The CY23FP12-002 is based on flash technology, so it can be reprogrammed up to 100 times. This enables fast and easy design changes and product updates, and eliminates any issues with old and out-of-date inventory.
Samples and small prototype quantities can be programmed on the CY3672-USB programmer. Cypress's value-added distribution partners and third-party programming systems from BP Microsystems, HiLo Systems, and others are available for large production quantities.

## CyberClocks ${ }^{\text {TM }}$ Software

CyberClocks is an easy-to-use software application that allows the user to custom-configure the CY23FP12-002. Users can specify the REF, PLL frequency, output frequencies and/or post-dividers, and different functional options. CyberClocks outputs an industry standard JEDEC file used for programming the CY23FP12-002.
CyberClocks can be downloaded free of charge from the Cypress website at www.cypress.com.

## CY3672-USB Development Kit

The Cypress CY3672-USB Developer Kit, in combination with the CY3692 Socket Adapter, is used to program samples and small prototype quantities of the CY23FP12-002. This portable programmer connects to a PC through a USB interface.
The JEDEC file output of CyberClocks can be downloaded to the portable programmer for small-volume programming, or for use with a production programming system for larger volumes.

## CY23FP12-002 Frequency Calculation

The CY23FP12-002 is an extremely flexible clock buffer with up to 12 individual outputs, generated from an integrated PLL. Four variables are used to determine the final output frequency. These are the input reference frequency, the M and N dividers, and the post divider.
The basic PLL block diagram is shown in Figure 2 on page 4. Each of the six clock output pairs has many post divider options available to it. $X$ is a programmable value between 5 and 130, and $2 X$ is twice that value. There are six post divider options: /1, $/ 2, / 3, / 4, / \mathrm{X}$, and /2X. The post divider options can be applied to the calculated PLL frequency or to the REF directly. The feedback is connected either internally to CLKA0 or externally to any output.
A programmable divider, $M$, is inserted between the reference input, REF, and the phase detector. The divider $M$ can be any integer 1 to 256. The PLL input frequency cannot be lower than 10 MHz or higher than 200 MHz . A programmable divider, N , is inserted between the feedback input, FBK, and the phase detector. The divider N can be any integer 1 to 256 . The PLL input frequency cannot be lower than 10 MHz or higher than 200 MHz .
The output can be calculated as follows:
$\mathrm{F}_{\mathrm{REF}} / \mathrm{M}=\mathrm{F}_{\mathrm{FBK}} / \mathrm{N}$.
$\mathrm{F}_{\mathrm{PLL}}=\left(\mathrm{F}_{\mathrm{REF}}{ }^{*} \mathrm{~N} *\right.$ post divider $) / \mathrm{M}$.
$\mathrm{F}_{\text {OUT }}=\mathrm{F}_{\mathrm{PLL}} /$ post divider.
In addition to above divider options, another option bypasses the PLL and passes the REF directly to the output.
$F_{\text {OUT }}=F_{\text {REF }}$

## Absolute Maximum Conditions

Exceeding maximum ratings may shorten the useful life of the device. User guidelines are not tested.

| Parameter | Description | Condition | Min. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{DD}}$ | Supply voltage | Nonfunctional | -0.5 | 7 | VDC |
| $\mathrm{V}_{\text {IN }}$ | Input voltage REF | Relative to $\mathrm{V}_{\mathrm{CC}}$ | -0.5 | $\mathrm{V}_{\mathrm{DD}}+0.5$ | VDC |
| $\mathrm{V}_{\text {IN }}$ | Input voltage Except REF | Relative to $\mathrm{V}_{\mathrm{CC}}$ | -0.5 | $\mathrm{V}_{\mathrm{DD}}+0.5$ | VDC |
| LU | Latch up immunity | Functional | 300 |  | mA |
| $\mathrm{T}_{\mathrm{S}}$ | Temperature, storage | Nonfunctional | -65 | 125 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{J}}$ | Junction temperature | - | - | 125 | ${ }^{\circ} \mathrm{C}$ |
| $\varnothing_{\text {Jc }}$ | Dissipation, junction to case | Functional | 30 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\varnothing_{\text {Ja }}$ | Dissipation, junction to ambient | Functional | 67 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $E S D_{\text {h }}$ | ESD protection (human body model) | - | 2000 |  | V |
| $\mathrm{M}_{\text {SL }}$ | Moisture sensitivity level | - | MSL - 1 |  | class |
| $\mathrm{G}_{\text {ATES }}$ | Total functional gate count | Assembled die | 21375 |  | each |
| UL-94 | Flammability rating | At 1/8 in. | V-0 |  | class |
| FIT | Failure in time | Manufacturing test | 10 |  | ppm |

## Operating Conditions

| Parameter | Description | Test Conditions | Min. | Max. | Unit |
| :--- | :--- | :--- | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{DDC}}$ | Core supply voltage | - | 3.135 | 3.465 | V |
| $\mathrm{~V}_{\mathrm{DDA}}, \mathrm{V}_{\mathrm{DDB}}$ | Bank A, bank B supply voltage | - | 3.135 | 3.465 | V |
|  |  | - | 2.375 | 2.625 | V |
| $\mathrm{~T}_{\mathrm{A}}$ | Temperature, operating ambient | Commercial temperature | 0 | 70 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{t}_{\mathrm{PU}}$ | Power-up time for all $\mathrm{V}_{\mathrm{DD}}$ s to reach <br> minimum specified voltage (power ramps <br> must be monotonic) | - | 0.05 | 500 | ms |
|  |  |  |  |  |  |

## DC Electrical Specifications

| Parameter | Description | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {IL }}$ | Input LOW voltage ${ }^{[3]}$ | - | - | - | $0.3 \times V_{\text {DD }}$ | V |
| $\mathrm{V}_{\mathrm{IH}}$ | Input HIGH voltage ${ }^{[3]}$ | - | $0.7 \times V_{\text {DD }}$ | - | - | V |
| $\mathrm{I}_{\text {IL }}$ | Input LOW current ${ }^{[3]}$ | $\mathrm{V}_{\text {IN }}=0 \mathrm{~V}$ | - | - | 50 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\mathrm{H}}$ | Input HIGH current ${ }^{[3]}$ | $\mathrm{V}_{\text {IN }}=\mathrm{V}_{\mathrm{DD}}$ | - | - | 50 | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\text {OL }}$ | Output LOW voltage ${ }^{[4]}$ | $\mathrm{V}_{\mathrm{DDA}} / \mathrm{V}_{\mathrm{DDB}}=3.3 \mathrm{~V}, \mathrm{I}_{\mathrm{OL}}=16 \mathrm{~mA}$ (standard drive) <br> $\mathrm{V}_{\mathrm{DDA}} / \mathrm{V}_{\mathrm{DDB}}=3.3 \mathrm{~V}, \mathrm{I}_{\mathrm{OL}}=20 \mathrm{~mA}$ (high drive) <br> $\mathrm{V}_{\mathrm{DDA}} / \mathrm{V}_{\mathrm{DDB}}=2.5 \mathrm{~V}, \mathrm{I}_{\mathrm{OL}}=16 \mathrm{~mA}$ (high drive) | - | - | 0.5 | V |
| $\mathrm{V}_{\mathrm{OH}}$ | Output HIGH voltage ${ }^{[4]}$ | $\mathrm{V}_{\mathrm{DDA}} / \mathrm{V}_{\mathrm{DDB}}=3.3 \mathrm{~V}, \mathrm{I}_{\mathrm{OH}}=-16 \mathrm{~mA}$ (standard drive) <br> $\mathrm{V}_{\mathrm{DDA}} / \mathrm{V}_{\mathrm{DDB}}=3.3 \mathrm{~V}, \mathrm{I}_{\mathrm{OH}}=-20 \mathrm{~mA}$ (high drive) <br> $\mathrm{V}_{\mathrm{DDA}} / \mathrm{V}_{\mathrm{DDB}}=2.5 \mathrm{~V}, \mathrm{I}_{\mathrm{OH}}=-16 \mathrm{~mA}$ (high drive) | $V_{D D}-0.5$ | - | - | V |
| IDDS | Power-down supply current | REF $=0 \mathrm{MHz}$ | - | 12 | 50 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\mathrm{DD}}$ | Supply current | $\mathrm{V}_{\text {DDA }}=\mathrm{V}_{\text {DDB }}=2.5 \mathrm{~V}$, Unloaded outputs at 166 MHz | - | 40 | 65 | mA |
|  |  | $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDB}}=2.5 \mathrm{~V}$, Loaded outputs at 166 MHz , $C_{L}=15 \mathrm{pF}$ | - | 65 | 100 |  |
|  |  | $\mathrm{V}_{\text {DDA }}=\mathrm{V}_{\text {DDB }}=3.3 \mathrm{~V}$, Unloaded outputs at 166 MHz | - | 50 | 80 |  |
|  |  | $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDB}}=3.3 \mathrm{~V}$, Loaded outputs at $166 \mathrm{MHz}, \mathrm{C}_{\mathrm{L}}$ $=15 \mathrm{pF}$ | - | 100 | 120 |  |
| Notes <br> 3. Applies to both REF Clock and FBK. <br> 4. Parameter is guaranteed by design and characterization. Not $100 \%$ tested in production. |  |  |  |  |  |  |

## Switching Characteristics ${ }^{[5]}$

| Parameter | Description | Test Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\text {REF }}$ | Reference frequency ${ }^{[6]}$ |  | 10 | - | 200 | MHz |
| $\mathrm{ER}_{\text {REF }}$ | Reference edge rate |  | 1 | - | - | V/ns |
| DC ${ }_{\text {REF }}$ | Reference duty cycle |  | 25 | - | 75 | \% |
| fout | Output frequency ${ }^{[7]}$ | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$ | 10 | - | 200 | MHz |
|  |  | $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}$ | 10 | - | 100 |  |
| DC ${ }_{\text {OUT }}$ | Output duty cycle ${ }^{[5]}$ | $\mathrm{V}_{\text {DDA/B }}=3.3 \mathrm{~V}$, measured at $\mathrm{V}_{\mathrm{DD}} / 2$ | 45 | 50 | 55 | \% |
|  |  | $\mathrm{V}_{\text {DDA/B }}=2.5 \mathrm{~V}$, measured at $\mathrm{V}_{\mathrm{DD}} / 2$ | 40 | 50 | 60 |  |
| $\mathrm{t}_{3}$ | Rise time ${ }^{[5]}$ | $\mathrm{V}_{\mathrm{DDA} / \mathrm{B}}=3.3 \mathrm{~V}, 0.8 \mathrm{~V}$ to 2.0 V , <br> $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}$ (standard drive and high drive) | - | - | 1.6 | ns |
|  |  | $\mathrm{V}_{\mathrm{DDA} / \mathrm{B}}=3.3 \mathrm{~V}, 0.8 \mathrm{~V}$ to 2.0 V , <br> $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$ (standard drive and high drive) | - | - | 0.8 |  |
|  |  | $\mathrm{V}_{\mathrm{DDA} / \mathrm{B}}=2.5 \mathrm{~V}, 0.6 \mathrm{~V}$ to 1.8 V , $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF}$ (high drive only) | - | - | 2.0 |  |
|  |  | $\mathrm{V}_{\mathrm{DDA} / \mathrm{B}}=2.5 \mathrm{~V}, 0.6 \mathrm{~V}$ to 1.8 V , $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$ (high drive only) | - | - | 1.0 |  |
| $\mathrm{t}_{4}$ | Fall time ${ }^{[5]}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{DDA} / \mathrm{B}}=3.3 \mathrm{~V}, 0.8 \mathrm{~V} \text { to } 2.0 \mathrm{~V}, \\ & \mathrm{C}_{\mathrm{L}}=30 \mathrm{pF} \text { (standard drive and high drive) } \end{aligned}$ | - | - | 1.6 | ns |
|  |  | $\mathrm{V}_{\mathrm{DDA} / \mathrm{B}}=3.3 \mathrm{~V}, 0.8 \mathrm{~V} \text { to } 2.0 \mathrm{~V},$ <br> $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$ (standard drive and high drive) | - | - | 0.8 |  |
|  |  | $\mathrm{V}_{\mathrm{DDA} / \mathrm{B}}=2.5 \mathrm{~V}, 0.6 \mathrm{~V} \text { to } 1.8 \mathrm{~V} \text {, }$ $\mathrm{C}_{\mathrm{L}}=30 \mathrm{pF} \text { (high drive only) }$ | - | - | 1.6 |  |
|  |  | $\mathrm{V}_{\mathrm{DDA} / \mathrm{B}}=2.5 \mathrm{~V}, 0.6 \mathrm{~V}$ to 1.8 V , $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$ (high drive only) | - | - | 0.8 |  |
| TTB | Total timing budget, ${ }^{[8,9]}$ Bank $A$ and $B$ same frequency | Outputs at 200 MHz , tracking skew not included | - | - | 650 | ps |
|  | Total timing budget, bank A and $B$ different frequency |  | - | - | 850 |  |
| $\mathrm{t}_{5}$ | Output-to-output skew ${ }^{[5]}$ | All outputs equally loaded | - | - | 200 | ps |
|  | Bank-to-bank skew | Same frequency | - | - | 200 |  |
|  | Bank-to-bank skew | Different frequency | - | - | 400 |  |
|  | Bank-to-bank skew | Different voltage, same frequency | - | - | 400 |  |
| $\mathrm{t}_{6}$ | Input-to-output skew (static phase offset) ${ }^{[5]}$ | Measured at $\mathrm{V}_{\mathrm{DD}} / 2$, REF to FBK | - | 0 | 250 | ps |
| $\mathrm{t}_{7}$ | Device-to-device skew ${ }^{[5]}$ | Measured at $\mathrm{V}_{\mathrm{DD}} / 2$ | - | 0 | 500 | ps |
| $\mathrm{t}_{\mathrm{J}}$ | Cycle-to-cycle jitter ${ }^{[5]}$ (Peak-to-peak) | Bank A and B same frequency | - | - | 200 | ps |
|  | Cycle-to-cycle jitter ${ }^{[5]}$ (Peak-to-peak) | Bank A and B different frequency | - | - | 400 |  |

## Notes

5. All parameters are specified with loaded outputs.
6. When the device is configured as a non-PLL fanout buffer (PLL Power down enabled), the reference frequency can be lower than 10 MHz . With auto power down disabled and PLL power down enabled, the reference frequency can be as low as DC level.
7. When the device is configured as a non-PLL fanout buffer (PLL Power down enabled), the output frequency can be lower than 10 MHz . With auto power down disabled and PLL power down enabled, the output frequency can be as low as DC level.
8. Guaranteed by statistical correlation. Tested initially and after any design or process changes that may affect these parameters.
9. TTB is the window between the earliest and the latest output clocks with respect to the input reference clock across variations in output frequency, supply voltage, operating temperature, input clock edge rate, and process. The measurements are taken with the AC test load specified and include output-output skew, cycle-cycle jitter, and dynamic phase error.TTB will be equal to or smaller than the maximum specified value at a given frequency.

Switching Characteristics (continued) ${ }^{[5]}$

| Parameter | Description | Test Conditions | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{T} \text { SK }}$ | Tracking skew | Input reference clock at < 50-KHz modulation <br> with $\pm 3.75 \% ~ s p r e a d ~$ | - | - | 200 | ps |
| $\mathrm{t}_{\text {LOCK }}$ | PLL lock time ${ }^{[5]}$ | Stable power supply, valid clock at REF | - | - | 1.0 | ms |
| $\mathrm{t}_{\text {LD }}$ | Inserted loop delay | Max loop delay for PLL Lock (stable frequency $)$ | - | - | 7 | ns |
|  |  | Max loop delay to meet Tracking Skew Spec | - | - | 4 | ns |

## Switching Waveforms

Figure 3. Duty Cycle Timing


Figure 4. All Outputs Rise/Fall Time


Figure 5. Output-Output Skew


Figure 6. Input-Output Propagation Delay


Figure 7. Device-Device Skew


## Test Circuits



Test Circuit for all parameters

## Ordering Information

| Ordering Code | Package Type | Operating Range |
| :--- | :--- | :--- |
| Pb-free | 28-pin SSOP | Commercial, $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ |
| CY23FP12OXC-002 | 28-pin SSOP - Tape and Reel | Commercial, $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ |
| CY23FP12OXC-002T | Programmer with USB Interface |  |
| Programmer | CY23FP12 Socket Adapter for CY3672-USB Programmer (Label CY3672 ADP006) |  |
| CY3672-USB | CY3692 |  |

## Ordering Code Definition



## Package Drawing and Dimension

Figure 8. 28 -Pin ( 5.3 mm ) Shrunk Small Outline Package


## Acronyms

| Acronym | Description |
| :--- | :--- |
| DCXO | digitally controlled crystal oscillator |
| ESD | electrostatic discharge |
| PLL | phase-locked loop |
| RMS | root mean square |
| SSOP | shrunk small outline package |
| XTAL | crystal |

## Document Conventions

## Units of Measure

| Symbol | Unit of Measure |
| :--- | :--- |
| ${ }^{\circ} \mathrm{C}$ | degree Celsius |
| $\mu \mathrm{A}$ | micro amperes |
| mA | milli amperes |
| ms | milli seconds |
| MHz | Mega Hertz |
| ns | nano seconds |
| pF | pico Farad |
| ps | pico seconds |
| V | Volts |

## Document History Page

Document Title: CY23FP12-002 200-MHz Field Programmable Zero Delay Buffer Document Number: 38-07644

| REV. | ECN NO. | Submission <br> Date | Orig. of <br> Change | Description of Change |
| :---: | :---: | :---: | :---: | :--- |
| $* *$ | 206761 | See ECN | RGL | New Data Sheet |
| *A | 2865396 | $01 / 25 / 2010$ | KVM | Updated template. <br> Removed references to industrial temperature range <br> Added captions to tables 1-4. <br> Added Operating Conditions table. <br> Various edits to text. <br> Removed "FTG" from text about the CY3672 programmer. <br> Added part numbers CY23FP12OXC-002, CY23FP12OXC-002T <br> Removed part numbers CY23FP12OC-002, CY23FP12OC-002T, <br> CY23FP12OI-002 and CY23FP12OI-002T <br> Changed part number CY3672 to CY3672-USB. <br> Updated package drawing. |
| *B | 3146346 | $01 / 18 / 2011$ | BASH | Modified VIN max value from 7 to V VD +0.5 in Absolute Maximum Conditions. <br> Added Acronyms, Document Conventions, and Ordering Code Definition |
| *C | 3373869 | $09 / 20 / 2011$ | PURU | Added hyper link in page 7 <br> Updated Package Drawing and Dimension. <br> Updated Absolute Maximum Conditions Table. |

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